

# A new inversion method for NMR signal processing

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**Abstract**—We present a new, semi-analytic inversion method for nuclear magnetic resonance (NMR) log measurements. Our method represents multiwait-time measurements via short sums of exponentials. The resulting sparse  $T_2$  distribution requires fewer  $T_2$  relaxation times than present in linearized inversion methods. The  $T_1$  relaxation times, and corresponding amplitudes are estimated via convex optimization and a semi-analytic algorithm. We obtain an efficient way to represent the NMR data that can be utilized to estimate petrophysical properties and for compression in logging-while-drilling applications.

## I. INTRODUCTION

Nuclear magnetic resonance (NMR) logging tools indirectly measure the amount of hydrogen atoms in a geological formation which provides a way to infer about its porosity and permeability. Currently, NMR logging tools are the only available tools that provide information about pore geometry and disposition of fluids. In this regard, NMR tools are invaluable in determining the quality, production planning and development of a reservoir.

While advancements in logging-while-drilling (LWD) tool design and manufacturing improve reliability of real-time NMR measurements, transmission of the raw measured or processed data from downhole to uphole is still limited by the telemetry bandwidth. Compression algorithms are utilized to transmit either raw or processed echo trains or petrophysical measurements derived from the  $T_2$  inversion process [5], [4]. Readers interested in the physics of NMR measurements and related inverse problems are referred to [3].

Motivated by the compression problem for LWD, we have developed a new inversion method for NMR log data and applied it to compute efficient representations of Carr-Purcell-Meiboom-Gill (CPMG) echo decay train measurements. These representations only require a small number of relaxation times  $T_2$  and  $T_1$ , and corresponding amplitudes, thus reducing the amount of parameters transmitted uphole.

Linear inversion methods select in advance a fixed set of  $T_2$  and  $T_1$  relaxation times and compute, solving a linear system, the corresponding amplitudes  $a$  which are compressed for transmission uphole [3], [5], [4]. These methods yield many more parameters than indicated by physical considerations. In contrast, non-linear optimization-based methods seek to estimate a small set of parameters  $(a, T_1, T_2)$ s, albeit at a

higher computational cost [7], [9]. Unlike current NMR data inversion methods, our method does not require predefined  $T_2$  and  $T_1$  values, nor does it solve a large non-linear optimization problem. It is a semi-analytic inversion method that computes an approximate representation of the data in terms of a sparse set of parameters  $(a, T_1, T_2)$ . Using a common set of exponentials to represent the data, we obtain the  $T_2$  values which we use subsequently to compute the amplitudes  $a$  via convex optimization. Finally,  $T_1$  values are obtained in an analytic fashion by appropriate averaging. In our preliminary experiments, the proposed method provides a more efficient representation of the data than those generated by linearized methods. We expect that our method will prove computationally less demanding than non-linear optimization methods.

## II. NMR INVERSION PROBLEM

NMR logging tools typically acquire CPMG echo decay trains. Given  $N$  multiwait-time measured echo trains,  $M_n$ ,  $n = 1, \dots, N$ , each consisting of  $K_n$  echoes,  $M_n(k)$ ,  $k = 1, \dots, K_n$ , the NMR inversion problem is typically formulated as follows: find a set of positive parameters  $(a_j, T_{1,j}, T_{2,j})$ , such that the error sequences  $\epsilon_n$  in

$$M_n(k) = \sum_{j=1}^J a_j \left( 1 - e^{-\frac{T_{W,n}}{T_{1,j}}} \right) e^{-\frac{k T_E}{T_{2,j}}} + \epsilon_n(k), \quad (1)$$

are within the level of noise [3, Section 6.2]. Here  $T_{2,j}$  are the  $T_2$  relaxation times,  $a_j$  are the  $T_2$  amplitudes (which are the partial porosity of the pores),  $T_{1,j}$  are the corresponding  $T_1$  relaxation times (associated with the size of the pores),  $T_{W,n}$  is the  $n^{th}$  wait-time and  $T_E$  is the time sample between consecutive echoes, also referred to as the echo-spacing. The wait times  $T_{W,n}$  are positive and distinct and we assume that they are ordered as  $T_{W,1} > T_{W,2} > \dots > T_{W,N}$ .

The inversion problem (1) may be solved using linear [3] or non-linear [11], [7], [8], [9] methods but is always an ill-posed problem with non-unique solutions [3], [8]. To address this issue, the problem is usually approached by fixing specific  $T_1$  and  $T_2$  relaxation time values or imposing artificial bounds on them. Regularization factors that impose smoothness on the solution may be used as well. In contrast, our approach

takes advantage of the exponential nature of the model (1). In practice, our method determines a small number of terms  $J$  required to represent all echo trains while exploiting physical bounds on the ratios between the relaxation times  $T_1$  and  $T_2$ .

### III. PROPOSED NEW ALGORITHM FOR INVERSION

#### A. Step 1: Estimating $T_{2,j}$

The expected error  $\epsilon$  of an exponential fit

$$\left| M(k) - \sum_{j=1}^J w_j \gamma_j^k \right| < \epsilon, \quad k = 0, \dots, K \quad (2)$$

is governed by the decay of the singular values of a rectangular Hankel matrix  $\mathbf{M}$  of entries  $[\mathbf{M}]_{m,l} = M(m+l)$ ,  $0 \leq m \leq K-L$ ,  $0 \leq l \leq L$ . Here  $L \leq K/2$  is a parameter which overestimates the minimal number of terms  $J$ , which, for physical reasons, is a small number. Two solutions to the exponential fit problem via Hankel matrices are presented in [10], [6]. In the recent approach [1], [2], a square Hankel matrix is considered and the minimal number of terms  $J$  in (2) is directly related to the index of the singular value of  $\mathbf{M}$  closest to  $\epsilon$ . Here, singular values are sorted in decreasing order and normalized so that  $\sigma_1 = 1$ .

Even though, for fixed  $n$ , the model (1) can be expressed in the form (2), where  $w_j$  is replaced by  $a_j \left(1 - e^{-\frac{T_{W,n}}{T_{1,j}}}\right)$  and  $\gamma_j$  by  $e^{-\frac{T_E}{T_{2,j}}}$ , the inversion problem requires determination of  $\gamma_j$ 's that can simultaneously fit the  $N$  echo trains  $M_n$ . We achieve this fit by performing a singular value decomposition of the matrix

$$\mathbf{M} = \begin{bmatrix} \frac{1}{\sqrt{K_1-L+1}} \mathbf{M}_1 \\ \frac{1}{\sqrt{K_2-L+1}} \mathbf{M}_2 \\ \vdots \\ \frac{1}{\sqrt{K_n-L+1}} \mathbf{M}_N \end{bmatrix},$$

where  $\mathbf{M}_n$  are Hankel matrices of entries  $[\mathbf{M}_n]_{m,l} = M_n(m+l)$ ,  $0 \leq l \leq L \leq \min_n \{K_n\}/2$ ,  $0 \leq j \leq K_n - L$ . We pick a singular value  $\sigma$  of  $\mathbf{M}$  close to the standard deviation of the errors  $\epsilon_n$ , i.e.  $E[\sum_n \epsilon_n^2]^{1/2}$  and form the polynomial of degree  $L-1$  whose coefficients are the entries of the right singular vector associated with  $\sigma$ . Due to the real positivity constraint on  $T_{2,j}$ , we set  $\gamma_j$  to be the roots of this polynomial that lie within  $[0, 1]$  and estimate  $T_{2,j} = T_E / \ln \gamma_j^{-1}$ . If the level of noise is too high or it is hard to estimate, we simply compute the roots in  $(0, 1)$  associated to all the right singular vectors of  $\mathbf{M}$  and pick the set that provides the best fit of the model (1). In addition, linear inversion methods could be used as a preliminary step to denoise the echo trains.

#### B. Step 2: Estimating $a_j$

To match (1), using the values  $\gamma_j$  of step 1, we solve a constrained non-negative least square problem

$$M_n(k) \approx \sum_{j=1}^J w_{n,j} \gamma_j^k$$

for  $w_{n,j}$ , with constraints  $w_{n,j} > 0$  and  $w_{n+1,j} < w_{n,j}$ , for  $n = 1, \dots, N-1$  which follow from the ordering of wait times. We show next that such a solution may be factor as

$$w_{n,j} = a_j p_{n,j}, \quad (3)$$

where  $a_j > 0$  and  $p_{n,j}$  are the polarization factors

$$p_{n,j} = 1 - e^{-\frac{T_{W,n}}{T_{1,j}}}. \quad (4)$$

In order to estimate  $p_{n,j}$  from  $w_{n,j}$ , let  $\beta_n \in (0, 1)$  be

$$\beta_n = T_{W,n+1}/T_{W,n}, \quad n = 1, \dots, N-1 \quad (5)$$

and observe that

$$\begin{aligned} \left(1 - \frac{w_{n,j}}{a_j}\right)^{\beta_n} &= (1 - p_{n,j})^{\beta_n} = e^{-\frac{\beta_n T_{W,n}}{T_{1,j}}} \\ &= e^{-\frac{T_{W,n+1}}{T_{1,j}}} = 1 - \frac{w_{n+1,j}}{a_j}, \end{aligned} \quad (6)$$

where we have used (3), (4), and (5). We rewrite (6) as

$$0 = y_{n,j}^{\beta_n} - q_{n,j} y_{n,j} + q_{n,j} - 1, \quad (7)$$

where  $q_{n,j} = w_{n+1,j}/w_{n,j}$  and  $y_{n,j} = 1 - p_{n,j}$  are both in  $(0, 1)$ . Thus, finding the polarization factors  $p_{n,j}$  is equivalent to finding zeros of  $g(y) = y^{\beta_n} - q_{n,j}y + q_{n,j} - 1$  for  $y \in (0, 1)$ . Note that

$$q_{n,j} = \frac{p_{n+1,j}}{p_{n,j}} = \frac{1 - e^{-\frac{T_{W,n+1}}{T_{1,j}}}}{1 - e^{-\frac{T_{W,n}}{T_{1,j}}}} > \frac{T_{W,n+1}}{T_{W,n}} = \beta_n, \quad (8)$$

which, together with  $\beta_n \in (0, 1)$ , implies that  $g$  is a strictly concave function on  $(0, 1)$  which attains its maximum at  $Y_{n,j} = (\beta_n/q_{n,j})^{\frac{1}{1-\beta_n}} < 1$ . Since  $g(0) = q_{n,j} - 1 < 0$ ,  $g(1) = 0$ , and  $g$  is strictly increasing in  $(0, Y_{n,j})$ , it has exactly one zero in  $(0, Y_{n,j})$ . Hence, for each  $n$ , (7) has a unique solution  $y_{n,j}$  and we set  $p_{n,j} = 1 - y_{n,j}$ . Due to (3), we estimate  $a_j$  as a weighted arithmetic mean

$$a_j \approx \sum_{n=1}^{N-1} \frac{w_{n,j}}{p_{n,j}} P_a(n), \quad (9)$$

where the probability measure  $P_a$  (see Section III-D) excludes very small values of  $p_{n,j}$  generated when  $T_{W,n}/T_{1,j}$  is very small. Also, if  $T_{W,n}/T_{1,j}$  is large, the polarization factor  $p_{n,j}$  is very close to 1 and we can use (3) to directly estimate  $a_j$  as  $w_{n,j}$ .

#### C. Step 3: Estimating $T_{1,j}$

Using (9), we introduce the new estimate  $p_{n,j} = w_{n,j}/a_j$  which, by (4), yields

$$\frac{1}{T_{1,j}} = -\frac{1}{T_{W,n}} \ln \left(1 - \frac{w_{n,j}}{a_j}\right), \quad (10)$$

for each  $n$ . Similar to [12], [13], where the expectation of  $T_1$  relaxation times are computed via a harmonic mean based on

TABLE I  
MEASUREMENT ( $a_j, \chi_j, T_{2,j}$ ) AND ACQUISITION ( $T_E, T_{W,n}, K_n$ )  
PARAMETERS USED TO GENERATE THE SYNTHETIC DATA.

$j$	$a_j$	$\chi_j = T_{1,j}/T_{2,j}$	$T_{2,j}$ (s)	$n$	$T_{W,n}$ (s)	$K_n$
1	0.0411	1.25	0.0224	1	9.0	1000
2	0.0412	1.25	0.0259	2	3.0	1000
3	0.0391	1.25	0.0300	3	1.0	1000
4	0.0011	2	1.1589	4	0.3	300
5	0.0260	2	1.3413	5	0.1	100
$T_E = 1$ ms						
				6	0.03	30
				7	0.01	10

known distributions of  $T_2$  relaxation times, we estimate the corresponding  $T_{1,j}$  as a weighted harmonic mean:

$$T_{1,j} = \left[ -\sum_{n=1}^N \frac{1}{T_{W,n}} \ln \left( 1 - \frac{w_{n,j}}{a_j} \right) P_{T_1}(n) \right]^{-1},$$

for an appropriately chosen probability measure  $P_{T_1}$ , which we discuss next.

#### D. A choice for the probability measures $P_a$ and $P_{T_1}$

As already pointed out, for long wait times, (4) implies  $p_{n,j} \approx 1$  and, hence,  $w_{n,j}$  is already a good estimate for  $a_j$ . On the other hand, short wait times provide better estimates for  $T_{1,j}$ . In our numerical examples, we choose a uniform distribution for  $P_\alpha$  ( $\alpha$  is either  $a$  or  $T_1$ ) defined as

$$P_\alpha(n) = \frac{1}{|I_\alpha|} \sum_{m \in I_\alpha} \delta_{n,m}$$

for some index set  $I_\alpha \subset \{1, \dots, N\}$  having  $|I_\alpha|$  number of elements, where  $\delta_{m,n}$  is the Kronecker delta function, equal to one when  $m = n$  and zero otherwise.  $I_a$  contains indices corresponding to long wait times and  $I_{T_1}$  indices corresponding to short wait times. In this way we avoid numerical errors that direct use of (10) could cause.

### IV. NUMERICAL EXAMPLES

#### A. Noise-free case

We test the proposed method on noise-free synthetic data generated using (1) with  $\epsilon_n(k) = 0$ , for all  $n, k$ . The acquisition and measurement parameters are listed in Table I. The synthetic data, its approximation, and the logarithm of the absolute error (which is less than  $10^{-8}$ ) are displayed in Figure 1. The relative errors of the estimated parameters are listed in Table II.

#### B. Noisy case

We also test the proposed algorithm on simulated noisy measurements by adding zero-mean Gaussian white noise with a standard deviation of 0.005 to the noise-free synthetic data shown in Figure 1. The noisy measurements, our denoised approximation (with  $J = 2$  terms), and their difference are displayed in Figure 2. This difference lies within the noise level.

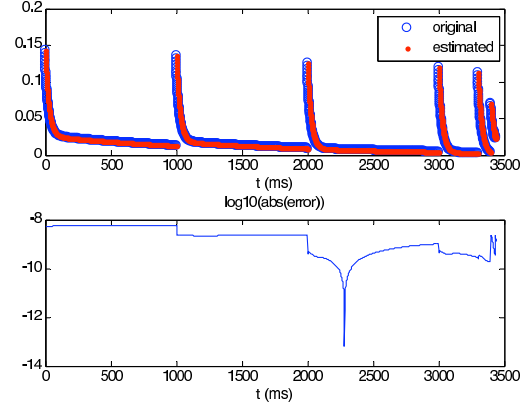


Fig. 1. [Top] Synthetic data (blue) generated using the parameters in Table I and the corresponding estimate using the proposed method (red). [Bottom] Logarithm of the approximation error.

TABLE II  
RELATIVE ERROR OF ESTIMATED PARAMETERS

$j$	$ \tilde{a}_j - a_j /a_j$	$ \tilde{\chi}_j - \chi_j /\chi_j$	$ \tilde{T}_{2,j} - T_{2,j} /T_{2,j}$
1	$6.9042 \times 10^{-7}$	$1.937 \times 10^{-6}$	$5.0104 \times 10^{-8}$
2	$1.4627 \times 10^{-6}$	$2.9483 \times 10^{-6}$	$2.7475 \times 10^{-7}$
3	$2.2767 \times 10^{-6}$	$1.2760 \times 10^{-6}$	$9.3993 \times 10^{-8}$
4	$9.3651 \times 10^{-3}$	$1.2843 \times 10^{-4}$	$6.4953 \times 10^{-4}$
5	$4.1239 \times 10^{-4}$	$6.3996 \times 10^{-6}$	$3.2651 \times 10^{-5}$

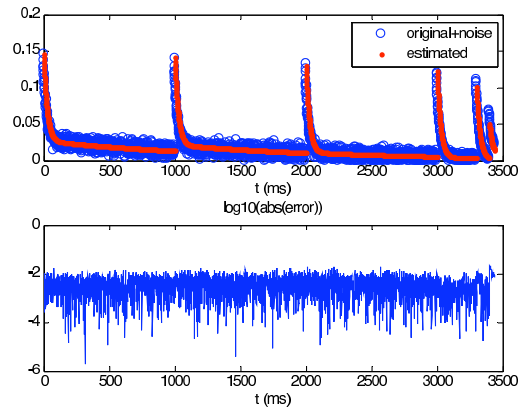


Fig. 2. [Top] Noisy measurement (blue) and denoised approximation (red). [Bottom] The logarithm of the absolute value of the difference between them.

TABLE III  
PARAMETERS ESTIMATED FROM NOISY VERSION OF SYNTHETIC DATA  
PRESENTED IN FIGURE 1.

$j$	$a_j$	$\chi_j = T_{1,j}/T_{2,j}$	$T_{2,j}$
1	0.1236	2.2184	$0.0255 \times 10^3$
2	0.0275	1.3973	$1.3685 \times 10^3$

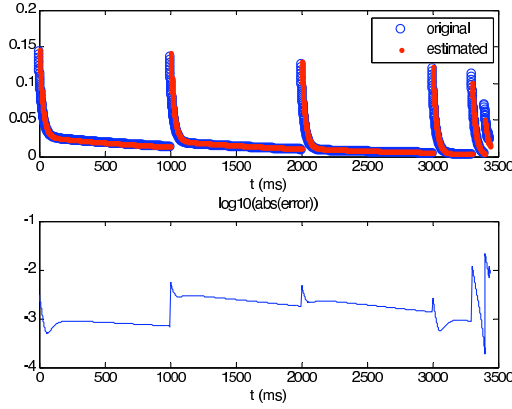


Fig. 3. [Top] Noise-free data (blue) and denoised approximation of the noisy data (red). [Bottom] The logarithm of the absolute value of the difference between them.

TABLE IV

ESTIMATE OF POROSITY,  $\Phi = \sum_{j=1}^J a_j$ , OBTAINED FROM NOISY DATA COMPARED WITH THE EXACT POROSITY OF THE SYNTHETIC NOISE-FREE DATA GENERATED USING THE PARAMETERS IN TABLE 3.

Porosity estimated from noise-free data (Table I)	Porosity estimated from noisy data (Table III)	Relative error
$\Phi_0 = 0.1485$	$\Phi_E = 0.1511$	$\frac{ \Phi_E - \Phi_0 }{\Phi_0} = 0.0175$

## V. DISCUSSION

In Figure 3 (top) we superimposed the denoised approximation of the data in Figure 2 with the approximation of the noise-free data (see Figure 1) generated by the parameters listed in Table I. The denoised approximation requires only  $J = 2$  terms for the error to stay within the noise level. Furthermore, when we compare the porosity computed using the amplitudes of the noise-free data and the denoised approximation, the relative error is less than 2% (see Table IV).

In practice, linearized inversion methods use between 16 and 32  $T_2$  relaxation times and between 1 and 5  $T_1$  relaxation times, hence, the need to compress and transmit up-hole in the range of 16 and 160 amplitudes. In our numerical experiments, we observed that at most 4, if not less, triples  $(a, T_1, T_2)$  were sufficient to achieve an approximation within the noise level. Therefore, only 12 values are compressed and transmitted up-hole. Compared to the best linearized inversion scenario, the proposed method provides, at least, a 25% reduction in the number of parameters compressed and transmitted up-hole.

## VI. CONCLUSION

We have presented a new, semi-analytic inversion method for nuclear magnetic resonance log data. This method assumes sparsity on the  $T_2$  relaxation times and, consequently, finds a sparse model to represent the data within the noise level. The sparsity assumption eliminates the need for processing

parameters present in linearized inversion methods. Because our method is a semi-analytic method, it is potentially more efficient than non-linear optimization-based inversion methods.

The resulting  $T_1$  and  $T_2$  relaxation times and corresponding amplitudes are useful for the estimation of petrophysical properties and for data compression in LWD applications. For LWD applications, our method produces fewer values to be compressed and transmitted up-hole than linearized inversion methods.

## VII. ACKNOWLEDGMENTS

The authors thank the management of WesternGeco and Schlumberger for permission to publish this work. Funding for this research has been provided by Schlumberger - Houston Formation & Evaluation Principal Investigators' 2012 Innovation Award. The first author would like to thank Konstantin Osypov and Jadeiva Goswami for their continuous support during the progress of this work.

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