# Graph Signal Processing: Fundamentals and Applications to Diffusion Processes 

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## Network Science analytics



Internet


Clean energy and grid analytics


- Desiderata: Process, analyze and learn from network data [Kolaczyk'09]
- Network as graph $G=(\mathcal{V}, \mathcal{E})$ : encode pairwise relationships
- Interest here not in $G$ itself, but in data associated with nodes in $\mathcal{V}$ $\Rightarrow$ Object of study is a graph signal $\mathbf{x}$
- Q: Graph signals common and interesting as networks are?


## Network of economic sectors of the United States

- Bureau of Economic Analysis of the U.S. Department of Commerce
- $\mathcal{E}=$ Output of sector $i$ is an input to sector $j$ ( 62 sectors in $\mathcal{V}$ )

Oil and Gas


Finance


- Oil extraction (OG), Petroleum and coal products (PC), Construction (CO)
- Administrative services (AS), Professional services (MP)
- Credit intermediation (FR), Securities (SC), Real state (RA), Insurance (IC)
- Only interactions stronger than a threshold are shown


## Network of economic sectors of the United States

- Bureau of Economic Analysis of the U.S. Department of Commerce
- $\mathcal{E}=$ Output of sector $i$ is an input to sector $j$ ( 62 sectors in $\mathcal{V}$ )

- A few sectors have widespread strong influence (services, finance, energy)
- Some sectors have strong indirect influences (oil)
- The heavy last row is final consumption
- This is an interesting network $\Rightarrow$ Signals on this graph are as well


## Disaggregated GDP of the United States

- Signal $\mathbf{x}=$ output per sector $=$ disaggregated GDP
$\Rightarrow$ Network structure used to, e.g., reduce GDP estimation noise

- Signal is as interesting as the network itself. Arguably more
- Same is true on brain connectivity and fMRI brain signals, ...
- Gene regulatory networks and gene expression levels, ...
- Online social networks and information cascades, ...
- Alignment of customer preferences and product ratings, ...


## Graph signal processing

- Graph SP: broaden classical SP to graph signals [Shuman et al.'13] $\Rightarrow$ Our view: GSP well suited to study network (diffusion) processes

- As.: Signal properties related to topology of $G$ (locality, smoothness)
$\Rightarrow$ Algorithms that fruitfully leverage this relational structure
- Q: Why do we expect the graph structure to be useful in processing $\mathbf{x}$ ?


## Importance of signal structure in time

- Signal and Information Processing is about exploiting signal structure
- Discrete time described by cyclic graph
$\Rightarrow$ Time $n$ follows time $n-1$
$\Rightarrow$ Signal value $x_{n}$ similar to $x_{n-1}$
- Formalized with the notion of frequency

- Cyclic structure $\Rightarrow$ Fourier transform $\Rightarrow \tilde{\mathbf{x}}=\mathbf{F}^{H} \mathbf{x}$ $\square$
- Fourier transform $\Rightarrow$ Projection on eigenvector space of cycle


## Covariances and principal components

- Random signal with mean $\mathbb{E}[\mathbf{x}]=0$ and covariance $\mathbf{C}_{x}=\mathbb{E}\left[\mathbf{x x}^{H}\right]$
$\Rightarrow$ Eigenvector decomposition $\mathrm{C}_{\mathrm{x}}=\mathrm{V} \boldsymbol{\wedge} \mathbf{V}^{H}$
- Covariance matrix $\mathbf{C}_{x}$ is a graph
$\Rightarrow$ Not a very good graph, but still
- Precision matrix $\mathbf{C}_{x}^{-1}$ a common graph too
$\Rightarrow$ Conditional dependencies of Gaussian $\mathbf{x}$

- Covariance matrix structure $\Rightarrow$ Principal components (PCA) $\Rightarrow \tilde{\mathbf{x}}=\mathbf{V}^{H} \mathbf{x}$
- PCA transform $\Rightarrow$ Projection on eigenvector space of (inverse) covariance
- Q: Can we extend these principles to general graphs and signals?


## Graphs 101

- Formally, a graph $G$ (or a network) is a triplet $(\mathcal{V}, \mathcal{E}, W)$
- $\mathcal{V}=\{1,2, \ldots, N\}$ is a finite set of $N$ nodes or vertices
- $\mathcal{E} \subseteq \mathcal{V} \times \mathcal{V}$ is a set of edges defined as ordered pairs ( $n, m$ )
- Write $\mathcal{N}(n)=\{m \in \mathcal{V}:(m, n) \in \mathcal{E}\}$ as the in-neighbors of $n$
- $W: \mathcal{E} \rightarrow \mathbb{R}$ is a map from the set of edges to scalar values $w_{n m}$
- Represents the level of relationship from $n$ to $m$
- Often weights are strictly positive, $W: \mathcal{E} \rightarrow \mathbb{R}_{++}$
- Unweighted graphs $\Rightarrow w_{n m} \in\{0,1\}$, for all $(n, m) \in \mathcal{E}$
- Undirected graphs $\Rightarrow(n, m) \in \mathcal{E}$ if and only if $(m, n) \in \mathcal{E}$ and

$$
w_{n m}=w_{m n}, \text { for all }(n, m) \in \mathcal{E}
$$

## Graphs - examples

- Unweighted and directed graphs (e.g., time)

- $\mathcal{V}=\{0,1, \ldots, 23\}$
- $\mathcal{E}=\{(0,1),(1,2), \ldots,(22,23),(23,0)\}$
- $W:(n, m) \mapsto 1$, for all $(n, m) \in \mathcal{E}$
- Unweighted and undirected graphs (e.g., image)

> V $=\{1,2,3, \ldots, 9\}$
> $\mathcal{E}=\{(1,2),(2,3), \ldots,(8,9),(1,4), \ldots,(6,9)\}$
> $W:(n, m) \mapsto 1$, for all $(n, m) \in \mathcal{E}$



- Weighted and undirected graphs (e.g., covariance)
- $\mathcal{V}=\{1,2,3,4\}$
- $\mathcal{E}=\{(1,1),(1,2), \ldots,(4,4)\}=\mathcal{V} \times \mathcal{V}$
- $W:(n, m) \mapsto \sigma_{n m}=\sigma_{m n}$, for all $(n, m)$


## Adjacency matrix

- Algebraic graph theory: matrices associated with a graph $G$
$\Rightarrow$ Adjacency $\mathbf{A}$ and Laplacian $\mathbf{L}$ matrices
$\Rightarrow$ Spectral graph theory: properties of $G$ using spectrum of $\mathbf{A}$ or $\mathbf{L}$
- Given $G=(\mathcal{V}, \mathcal{E}, W)$, the adjacency matrix $\mathbf{A} \in \mathbb{R}^{N \times N}$ is

$$
A_{n m}= \begin{cases}w_{n m}, & \text { if }(n, m) \in \mathcal{E} \\ 0, & \text { otherwise }\end{cases}
$$

- Matrix representation incorporating all information about $G$
$\Rightarrow$ For unweighted graphs, positive entries represent connected pairs
$\Rightarrow$ For weighted graphs, also denote proximities between pairs


## Degree and $k$-hop neighbors

- If $G$ is unweighted and undirected, the degree of node $i$ is $|\mathcal{N}(i)|$
$\Rightarrow$ In directed graphs, have out-degree and an in-degree
- Using the adjacency matrix in the undirected case
$\Rightarrow$ For node $i: \operatorname{deg}(i)=\sum_{j \in \mathcal{N}(i)} A_{i j}=\sum_{j} A_{i j}$
$\Rightarrow$ For all $N$ nodes: $\mathbf{d}=\mathbf{A} 1 \rightarrow$ Degree matrix: $\mathbf{D}:=\operatorname{diag}(\mathbf{d})$
- Q: Can this be extended to $k$-hop neighbors? $\rightarrow$ Powers of $\mathbf{A}$
$\Rightarrow\left[\mathbf{A}^{k}\right]_{i j}$ non-zero only if there exists a path of length $k$ from $i$ to $j$
$\Rightarrow$ Support of $\mathbf{A}^{k}$ : pairs that can be reached in $k$ hops

$\left[A^{2}\right]_{3,5}$



## Laplacian of a graph

- Given undirected $G$ with $\mathbf{A}$ and $\mathbf{D}$, the Laplacian matrix $\mathbf{L} \in \mathbb{R}^{N \times N}$ is

$$
\mathbf{L}=\mathbf{D}-\mathbf{A}
$$

$\Rightarrow$ Equivalently, L can be defined element-wise as

$$
L_{i j}=\left\{\begin{array}{lc}
\operatorname{deg}(i), & \text { if } i=j \\
-w_{i j}, & \text { if }(i, j) \in \mathcal{E} \\
0, & \text { otherwise }
\end{array}\right.
$$

- Normalized Laplacian: $\mathcal{L}=\mathbf{D}^{-1 / 2} \mathbf{L D}^{-1 / 2}$ (we will focus on L )


$$
\mathbf{L}=\left[\begin{array}{cccc}
3 & -1 & 0 & -2 \\
-1 & 6 & -3 & -2 \\
0 & -3 & 4 & -1 \\
-2 & -2 & -1 & 5
\end{array}\right]
$$

## Spectral properties of the Laplacian

- Denote by $\lambda_{i}$ and $\mathbf{v}_{i}$ the eigenvalues and eigenvectors of $\mathbf{L}$
- $\mathbf{L}$ is positive semi-definite
$\Rightarrow \mathbf{x}^{T} \mathbf{L} \mathbf{x}=\frac{1}{2} \sum_{(i, j) \in \mathcal{E}} w_{i j}\left(x_{i}-x_{j}\right)^{2} \geq 0$, for all $\mathbf{x}$
$\Rightarrow$ All eigenvalues are nonnegative, i.e. $\lambda_{i} \geq 0$ for all $i$
- A constant vector $\mathbf{1}$ is an eigenvector of $\mathbf{L}$ with eigenvalue 0

$$
[\mathbf{L} 1]_{i}=\sum_{j \in \mathcal{N}(i)} w_{i j}(1-1)=0
$$

$\Rightarrow$ Thus, $\lambda_{1}=0$ and $\mathbf{v}_{1}=(1 / \sqrt{N}) \mathbf{1}$

- In connected graphs, it holds that $\lambda_{i}>0$ for $i=2, \ldots, N$
$\Rightarrow$ Multiplicity $\{\lambda=0\}=$ number of connected components


## Part I: Fundamentals

## Motivation and preliminaries

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Graph signals and the shift operator
Graph Fourier Transform (GFT)
Graph filters and network processes
Part II: Applications
Filter design for network operators
Sampling graph signals Blind identification of graph filters
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Concluding remarks

## Graph signals

- Consider graph $G=(\mathcal{V}, \mathcal{E}, W)$. Graph signals are mappings $x: \mathcal{V} \rightarrow \mathbb{R}$ $\Rightarrow$ Defined on the vertices of the graph (data tied to nodes)

Ex: Opinion profile, buffer congestion levels, neural activity, epidemic

- May be represented as a vector $\mathbf{x} \in \mathbb{R}^{N}$
$\Rightarrow x_{n}$ denotes the signal value at the $n$-th vertex in $\mathcal{V}$
$\Rightarrow$ Implicit ordering of vertices (same as in $\mathbf{A}$ or $\mathbf{L}$ )

- Data associated with links of $G \Rightarrow$ Use line graph of $G$


## Graph signals - Genetic profiles

- Graphs representing gene-gene interactions
$\Rightarrow$ Each node denotes a single gene (loosely speaking)
$\Rightarrow$ Connected if their coded proteins participate in same metabolism
- Genetic profiles for each patient can be considered as a graph signal
$\Rightarrow$ Signal on each node is 1 if mutated and 0 otherwise


Sample patient 1 with subtype 1

- To understand a graph signal, the structure of $G$ must be considered


## Graph-shift operator

- To understand and analyze $\mathbf{x}$, useful to account for G's structure
- Associated with $G$ is the graph-shift operator $\mathbf{S} \in \mathbb{R}^{N \times N}$
$\Rightarrow S_{i j}=0$ for $i \neq j$ and $(i, j) \notin \mathcal{E}$ (captures local structure in $G$ )
- S can take nonzero values in the edges of $G$ or in its diagonal


$$
\mathbf{S}=\left(\begin{array}{cccccc}
S_{11} & S_{12} & 0 & 0 & S_{15} & 0 \\
S_{21} & S_{22} & S_{23} & 0 & S_{25} & 0 \\
0 & S_{23} & S_{33} & S_{34} & 0 & 0 \\
0 & 0 & S_{43} & S_{44} & S_{45} & S_{46} \\
S_{51} & S_{52} & 0 & S_{54} & S_{55} & 0 \\
0 & 0 & 0 & S_{64} & 0 & S_{66}
\end{array}\right)
$$

- Ex: Adjacency A, degree D, and Laplacian $\mathbf{L}=\mathbf{D}$ - $\mathbf{A}$ matrices


## Relevance of the graph-shift operator

- Q: Why is S called shift? A: Resemblance to time shifts

- S will be building block for GSP algorithms (More soon)
$\Rightarrow$ Same is true in the time domain (filters and delay)



## Local structure of graph-shift operator

S represents a linear transformation that can be computed locally at the nodes of the graph. More rigorously, if $\mathbf{y}$ is defined as $\mathbf{y}=\mathbf{S x}$, then node $i$ can compute $y_{i}$ if it has access to $x_{j}$ at $j \in \mathcal{N}(i)$.

- Straightforward because $[\mathbf{S}]_{i j} \neq 0$ only if $i=j$ or $(j, i) \in \mathcal{E}$


$$
\Rightarrow\left(\begin{array}{l}
y_{1} \\
y_{2} \\
y_{3} \\
y_{4} \\
y_{5} \\
y_{6}
\end{array}\right)=\left(\begin{array}{cccccc}
S_{11} & S_{12} & 0 & 0 & S_{15} & 0 \\
S_{21} & S_{22} & S_{23} & 0 & S_{25} & 0 \\
\hline 0 & S_{32} & S_{33} & S_{34} & 0 & 0 \\
0 & 0 & S_{43} & S_{44} & S_{45} & S_{46} \\
S_{51} & S_{52} & 0 & S_{54} & S_{55} & 0 \\
0 & 0 & 0 & S_{64} & 0 & S_{66}
\end{array}\right)\left(\begin{array}{l}
x_{1} \\
x_{2} \\
x_{3} \\
x_{4} \\
x_{5} \\
x_{6}
\end{array}\right)
$$

- What if $\mathbf{y}=\mathbf{S}^{2} \mathbf{x}$ ?
$\Rightarrow$ Like powers of
A: neighborhoods
$\Rightarrow y_{i}$ found using values within 2-hops

$$
\left[\mathbf{S}^{2}\right]_{3,5}=S_{3,2} S_{2,5}+S_{3,4} S_{4,5}
$$



## Graph Fourier Transform

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## Discrete Fourier Transform (DFT)

- Let $\mathbf{x}$ be a temporal signal, its DFT is $\tilde{\mathbf{x}}=\mathbf{F}^{H} \mathbf{x}$, with $F_{k n}=\frac{1}{\sqrt{N}} e^{+j \frac{2 \pi}{N} k n}$
$\Rightarrow$ Equivalent description, provides insights
$\Rightarrow$ Oftentimes, more parsimonious (bandlimited)
$\Rightarrow$ Facilitates the design of SP algorithms: e.g., filters
- Many other transformations (orthogonal dictionaries) exist

ANALYSIS

projection matrix

SYNTHESIS


- Q: What transformation is suitable for graph signals?


## Graph Fourier Transform (GFT)

- Useful transformation? $\Rightarrow \mathbf{S}$ involved in generation/description of $\mathbf{x}$
$\Rightarrow$ Let $\mathbf{S}=\mathbf{V} \wedge \mathbf{V}^{-1}$ be the shift associated with $G$
- The Graph Fourier Transform (GFT) of $\mathbf{x}$ is defined as

$$
\tilde{\mathbf{x}}=\mathbf{V}^{-1} \mathbf{x}
$$

- While the inverse GFT (iGFT) of $\tilde{\mathbf{x}}$ is defined as

$$
x=V \tilde{x}
$$

$\Rightarrow$ Eigenvectors $\mathbf{V}=\left[\mathbf{v}_{1}, \ldots, \mathbf{v}_{N}\right]$ are the frequency basis (atoms)

- Additional structure
$\Rightarrow$ If $\mathbf{S}$ is normal, then $\mathbf{V}^{-1}=\mathbf{V}^{H}$ and $\tilde{x}_{k}=\mathbf{v}_{k}^{H} \mathbf{x}=\left\langle\mathbf{v}_{k}, \mathbf{x}\right\rangle$
$\Rightarrow$ Parseval holds, $\|\mathbf{x}\|^{2}=\|\tilde{\mathbf{x}}\|^{2}$
- GFT $\Rightarrow$ Projection on eigenvector space of shift operator S


## Is this a reasonable transform?

- Particularized to cyclic graphs $\Rightarrow$ GFT $\equiv$ Fourier transform
- Particularized to covariance matrices $\Rightarrow$ GFT $\equiv$ PCA transform
- But really, this is an empirical question. GFT of disaggregated GDP

- GFT transform characterized by a few coefficients
$\Rightarrow$ Notion of bandlimitedness: $\mathbf{x}=\sum_{k=1}^{K} \tilde{x}_{k} \mathbf{v}_{k}$
$\Rightarrow$ Sampling, compression, filtering, pattern recognition


## Eigenvalues as frequencies

- Columns of $\mathbf{V}$ are the frequency atoms: $\mathbf{x}=\sum_{k} \tilde{x}_{k} \mathbf{v}_{k}$
- Q: What about the eigenvalues $\lambda_{k}=\Lambda_{k k}$
$\Rightarrow$ When $\mathbf{S}=\mathbf{A}_{d c}$, we get $\lambda_{k}=e^{-j \frac{2 \pi}{N} k}$
$\Rightarrow \lambda_{k}$ can be viewed as frequencies!!
- In time, well-defined relation between frequency and variation
$\Rightarrow$ Higher $k \Rightarrow$ higher oscillations
$\Rightarrow$ Bounds on total-variation: $T V(\mathbf{x})=\sum_{n}\left(x_{n}-x_{n-1}\right)^{2}$


- Q: Does this carry over for graph signals?
$\Rightarrow$ No in general, but if $\mathbf{S}=\mathbf{L}$ there are interpretations for $\lambda_{k}$
$\Rightarrow\left\{\lambda_{k}\right\}_{k=1}^{N}$ will be very important when analyzing graph filters


## Interpretation of the Laplacian

- Consider a graph $G$, let $\mathbf{x}$ be a signal on $G$, and set $\mathbf{S}=\mathbf{L}$
$\Rightarrow \mathbf{y}=\mathbf{S} \mathbf{x}$ is now $\mathbf{y}=\mathbf{L x} \Rightarrow y_{i}=\sum_{j \in \mathcal{N}(i)} w_{i j}\left(x_{i}-x_{j}\right)$
$\Rightarrow j$-th term is large if $x_{j}$ is very different from neighboring $x_{i}$
$\Rightarrow y_{i}$ measures difference of $x_{i}$ relative to its neighborhood
- We can also define the quadratic form $\mathbf{x}^{\top} \mathbf{S} \mathbf{x}$

$$
\mathbf{x}^{T} \mathbf{L} \mathbf{x}=\frac{1}{2} \sum_{(i, j) \in \mathcal{E}} w_{i j}\left(x_{i}-x_{j}\right)^{2}
$$

$\Rightarrow \mathbf{x}^{\top} \mathbf{L x}$ quantifies the (aggregated) local variation of signal $\mathbf{x}$
$\Rightarrow$ Natural measure of signal smoothness w.r.t. $G$

- Q: Interpretation of frequencies $\left\{\lambda_{k}\right\}_{k=1}^{N}$ when $\mathbf{S}=\mathbf{L}$ ?
$\Rightarrow$ If $\mathbf{x}=\mathbf{v}_{k}$, we get $\mathbf{x}^{T} \mathbf{L} \mathbf{x}=\lambda_{k} \Rightarrow$ local variation of $\mathbf{v}_{k}$
$\Rightarrow$ Frequencies account for local variation, they can be ordered
$\Rightarrow$ Eigenvector associated with eigenvalue 0 is constant

Frequencies of the Laplacian

- Laplacian eigenvalue $\lambda_{k}$ accounts for the local variation of $\mathbf{v}_{k}$ $\Rightarrow$ Let us plot some of the eigenvectors of $\mathbf{L}$ (also graph signals)

Ex: gene network, $N=10, k=1, k=2, k=9$


Ex: smooth natural images, $N=2^{16}, k=2, \ldots, 6$


## Application: Cancer subtype classification

- Patients diagnosed with same disease exhibit different behaviors
- Each patient has a genetic profile describing gene mutations
- Would be beneficial to infer phenotypes from genotypes
$\Rightarrow$ Targeted treatments, more suitable suggestions, etc.
- Traditional approaches consider different genes to be independent
$\Rightarrow$ Not ideal, as different genes may affect same metabolism
- Alternatively, consider genetic network
$\Rightarrow$ Genetic profiles become graph signals on genetic network
$\Rightarrow$ We will see how this consideration improves subtype classification


## Genetic network

- Undirected and unweighted gene-to-gene interaction graph
- 2458 nodes are genes in human DNA related to breast cancer
- An edge between two genes represents interaction
$\Rightarrow$ Coded proteins participate in the same metabolic process
- Adjacency matrix of the gene-interaction network



## Genetic profiles

- Genetic profile of 240 women with breast cancer
$\Rightarrow 44$ with serous subtype and 196 with endometrioid subtype
$\Rightarrow$ Patient $i$ has an associated profile $\mathbf{x}_{i} \in\{0,1\}^{2458}$
- Mutations are very varied across patients
$\Rightarrow$ Some patients present a lot of mutations
$\Rightarrow$ Some genes are consistently mutated across patients

- Q: Can we use genetic profiles to classify patients across subtypes?


## Improving $k$-nearest neighbor classification

- Distance between genetic profiles $\Rightarrow d(i, j)=\left\|\mathbf{x}_{i}-\mathbf{x}_{j}\right\|_{2}$
$\Rightarrow N$-fold cross-validation error from k-NN classification

$$
k=3 \Rightarrow 13.3 \%, \quad k=5 \Rightarrow 12.9 \%, \quad k=7 \Rightarrow 14.6 \%
$$

- Q: Can we do any better using graph signal processing?
- Each genetic profile $\mathbf{x}_{i}$ is a graph signal on the genetic network
$\Rightarrow$ Look at the frequency components $\tilde{\mathbf{x}}_{i}$ using the GFT
$\Rightarrow$ Use as shift operator S the Laplacian of the genetic network

Example of signal $\mathrm{x}_{i}$


Frequency representation $\tilde{\mathbf{x}}_{i}$


## Distinguishing Power

- Define the distinguishing power of frequency $\mathbf{v}_{k}$ as

$$
D P\left(\mathbf{v}_{k}\right)=\left|\frac{\sum_{i: y_{i}=1} \tilde{\mathbf{x}}_{i}(k)}{\sum_{i} \mathbf{1}\left\{y_{i}=1\right\}}-\frac{\sum_{i: y_{i}=2} \tilde{\mathbf{x}}_{i}(k)}{\sum_{i} \mathbf{1}\left\{y_{i}=2\right\}}\right| / \sum_{i}\left|\tilde{\mathbf{x}}_{i}(k)\right|,
$$

- Normalized difference between the mean GFT coefficient for $\mathbf{v}_{k}$ $\Rightarrow$ Among patients with serous and endometrioid subtypes
- Distinguishing power is not equal across frequencies

- The distinguishing power defined is one of many proper heuristics


## Increasing accuracy by selecting the best frequencies

- Keep information in frequencies with higher distinguishing power $\Rightarrow$ Filter, i.e., multiply $\tilde{\mathbf{x}}_{i}$ by $\operatorname{diag}\left(\tilde{\mathbf{h}}^{p}\right)$ where

$$
\left[\tilde{\mathbf{h}}^{p}\right]_{k}= \begin{cases}1, & \text { if } D P\left(\mathbf{v}_{k}\right) \geq p \text {-th percentile of } D P \\ 0, & \text { otherwise }\end{cases}
$$

- Then perform inverse GFT to get the graph signal $\hat{\mathbf{x}}_{i}$



## Graph filters and network processes

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## Linear (shift-invariant) graph filter

- A graph filter $H: \mathbb{R}^{N} \rightarrow \mathbb{R}^{N}$ is a map between graph signals

Focus on linear filters
$\Rightarrow$ map represented by an
$N \times N$ matrix


DEF1: Polynomial in S of degree $L$, with coeff. $\mathbf{h}=\left[h_{0}, \ldots, h_{L}\right]^{T}$

$$
\mathbf{H}:=h_{0} \mathbf{S}^{0}+h_{1} \mathbf{S}^{1}+\ldots+h_{L} \mathbf{S}^{L}=\sum_{l=0}^{L} h_{l} \mathbf{S}^{\prime} \quad \text { [Sandryhaila13] }
$$

DEF2: Orthogonal operator in the frequency domain

$$
\mathbf{H}:=\mathbf{V} \operatorname{diag}(\tilde{\mathbf{h}}) \mathbf{V}^{-1}, \quad \tilde{h}_{k}=g\left(\lambda_{k}\right)
$$

- With $[\Psi]_{k, l}:=\lambda_{k}^{l-1}$, we have $\tilde{\mathbf{h}}=\Psi \mathbf{h} \Rightarrow$ Defs can be rendered equivalent $\Rightarrow$ More on this later, now focus on DEF1


## Graph filters as linear network operators

- DEF1 says $\mathbf{H}=\sum_{l=0}^{L} h_{l} \mathbf{S}^{\prime}$
- Suppose H acts on a graph signal $\mathbf{x}$ to generate $\mathbf{y}=\mathbf{H x}$
$\Rightarrow$ If we define $\mathbf{x}^{(/)}:=\mathbf{S}^{\prime} \mathbf{x}=\mathbf{S} \mathbf{x}^{(/-1)}$

$$
\mathbf{y}=\sum_{l=0}^{L} h_{l} \mathbf{x}^{(I)}
$$

$y$ is a linear combination of successive shifted versions of $\mathbf{x}$

- After introducing $\mathbf{S}$, we stressed that $\mathbf{y}=\mathbf{S x}$ can be computed locally
$\Rightarrow \mathbf{x}^{(l)}$ can be found locally if $\mathbf{x}^{(l-1)}$ is known
$\Rightarrow$ The output of the filter can be found in $L$ local steps
- A graph filter represents a linear transformation that
$\Rightarrow$ Accounts for local structure of the graph
$\Rightarrow$ Can be implemented distributedly in $L$ steps
$\Rightarrow$ Only requires info in L-neighborhood [Shuman13, Sandyhaila14]


## An example of a graph filter

- $\mathbf{x}=[-1,2,0,0,0,0]^{T}, \mathbf{h}=[1,1,0.5]^{T}, \mathbf{y}=\left(\sum_{l=0}^{L} h_{l} \mathbf{S}\right) \mathbf{x}=\sum_{l=0}^{L} h_{l} \mathbf{x}^{(l)}$


$$
\begin{aligned}
& \mathbf{S}=\mathbf{A}=\left(\begin{array}{cccccc}
0 & 1 & 0 & 0 & 1 & 0 \\
1 & 0 & 1 & 0 & 1 & 0 \\
0 & 1 & 0 & 1 & 0 & 0 \\
0 & 0 & 1 & 0 & 1 & 1
\end{array}\right) \quad \mathbf{y}=\sum_{l=0}^{L} h_{l} \mathbf{S}^{l} \mathbf{x}=\sum_{l=0}^{L} h_{l} \mathbf{x}^{(l)} \\
& \square \\
& \mathbf{y}=h_{0} \mathbf{x}^{(0)}+h_{1} \mathbf{x}^{(1)}+h_{2} \mathbf{x}^{(2)}
\end{aligned}
$$

Given $\mathbf{x}=[-1,2,0,0,0,0]^{T}$ and $\mathbf{h}=[1,1,0.5]^{T} \Rightarrow$ Find $\left\{\mathbf{x}^{(0)}, \mathbf{x}^{(1)}, \mathbf{x}^{(2)}\right\} \Rightarrow$ Find $\mathbf{y}$


## Frequency response of a graph filter

- Recalling that $\mathrm{S}=\mathbf{V} \wedge \mathbf{V}^{-1}$, we may write

$$
\mathbf{H}=\sum_{l=0}^{L} h_{l} \mathbf{S}^{\prime}=\sum_{l=0}^{L} h_{l} \mathbf{V} \boldsymbol{\Lambda}^{\prime} \mathbf{V}^{-1}=\mathbf{V}\left(\sum_{l=0}^{L} h_{l} \boldsymbol{\Lambda}^{\prime}\right) \mathbf{V}^{-1}
$$

- The application Hx of filter H to $\mathbf{x}$ can be split into three parts
$\Rightarrow \mathbf{V}^{-1}$ takes signal $\mathbf{x}$ to the graph frequency domain $\tilde{\mathbf{x}}$
$\Rightarrow \tilde{\mathbf{H}}:=\sum_{l=0}^{L} h_{l} \boldsymbol{\Lambda}^{\prime}$ modifies the frequency coefficients to obtain $\tilde{\mathbf{y}}$
$\Rightarrow \mathbf{V}$ brings the signal $\tilde{\mathbf{y}}$ back to the graph domain $\mathbf{y}$
- Since $\tilde{\mathbf{H}}$ is diagonal, define $\tilde{\mathbf{H}}=: \operatorname{diag}(\tilde{\mathbf{h}})$
$\Rightarrow \tilde{\mathbf{h}}$ is the frequency response of the filter $\mathbf{H}$
$\Rightarrow$ Output at frequency $k$ depends only on input at frequency $k$

$$
\tilde{y}_{k}=\tilde{h}_{k} \tilde{x}_{k}
$$

## Frequency response and filter coefficients

- Relation between $\tilde{\mathbf{h}}$ and $\mathbf{h}$ in a more friendly manner?
$\Rightarrow$ Since $\tilde{\mathbf{h}}=\operatorname{diag}\left(\sum_{l=0}^{L} h_{l} \Lambda^{\prime}\right)$, we have that $\tilde{h}_{k}=\sum_{l=0}^{L} h_{l} \lambda_{k}^{\prime}$
$\Rightarrow$ Define the Vandermonde matrix $\Psi$ as

$$
\Psi:=\left(\begin{array}{cccc}
1 & \lambda_{1} & \ldots & \lambda_{1}^{L} \\
\vdots & \vdots & & \vdots \\
1 & \lambda_{N} & \ldots & \lambda_{N}^{L}
\end{array}\right)
$$

Frequency response of a graph filter
If $h$ are the coefficients of a graph filter, its frequency response is

$$
\tilde{\mathbf{h}}=\psi \mathbf{h}
$$

- Given a desired $\tilde{\mathbf{h}}$, we can find the coefficients $\mathbf{h}$ as

$$
\mathbf{h}=\Psi^{-1} \tilde{\mathbf{h}}
$$

$\Rightarrow$ Since $\Psi$ is Vandermonde, invertible as long as $\lambda_{k} \neq \lambda_{k^{\prime}}$ for $k \neq k^{\prime}$

## More on the frequency response

- Since $\mathbf{h}=\Psi^{-1} \tilde{\mathbf{h}} \Rightarrow$ If all $\left\{\lambda_{k}\right\}_{k=1}^{N}$ distinct, then
$\Rightarrow$ Any $\tilde{\mathbf{h}}$ can be implemented with at most $L+1=N$ coefficients
- Since $\mathbf{h}=\Psi \tilde{\mathbf{h}} \Rightarrow \operatorname{If} \lambda_{\boldsymbol{k}}=\lambda_{k^{\prime}}$, then
$\Rightarrow$ The corresponding frequency response will be the same $\tilde{h}_{k}=\tilde{h}_{k^{\prime}}$
- For the particular case when $\mathbf{S}=\mathbf{A}_{d c}$, we have that $\lambda_{k}=e^{-j \frac{2 \pi}{N}(k-1)}$

$$
\Psi=\left(\begin{array}{cccc}
1 & 1 & \cdots & 1 \\
1 & e^{-j \frac{2 \pi(1)(1)}{N}} & \cdots & e^{-j \frac{2 \pi(1)(N-1)}{N}} \\
\vdots & \vdots & & \vdots \\
1 & e^{-j \frac{2 \pi(N-1)(1)}{N}} & \ldots & e^{-j \frac{2 \pi(N-1)(N-1)}{N}}
\end{array}\right)=F^{H}
$$

$\Rightarrow$ The frequency response is the DFT of the impulse response

$$
\tilde{\mathbf{h}}=F^{H} \mathbf{h}
$$

## Frequency response for graph signals and filters

- Suppose that we have a signal $\mathbf{x}$ and filter coefficients $\mathbf{h}$
- For time signals, it holds that the output $\mathbf{y}$ is

$$
\tilde{\mathbf{y}}=\operatorname{diag}\left(\mathbf{F}^{H} \mathbf{h}\right) \mathbf{F}^{H} \mathbf{x}
$$

- For graph signals, the output $\mathbf{y}$ in the frequency domain is

$$
\tilde{\mathbf{y}}=\operatorname{diag}(\Psi \mathbf{h}) \mathbf{V}^{-1} \mathbf{x}
$$

- The GFT for filters is different from the GFT for signals $\Rightarrow$ Symmetry is lost, but both depend on spectrum of $\mathbf{S}$
$\Rightarrow$ Many of the properties are not true for graphs
$\Rightarrow$ Several options to generalize operations


## System identification and impulse response

- Suppose that our goal is to find $\mathbf{h}$ given $\mathbf{x}$ and $\mathbf{y}$
$\Rightarrow$ Using the previous expressions

$$
\mathbf{h}=\boldsymbol{\Psi}^{-1} \operatorname{diag}^{-1}\left(\mathbf{V}^{-1} \mathbf{x}\right) \mathbf{V}^{-1} \mathbf{y}
$$

- In time, if we set $\mathbf{x}=[1,0, \ldots, 0]^{T}=\mathbf{e}_{1}$ (i.e., $\tilde{\mathbf{x}}=\mathbf{1}$ ), we have $\Rightarrow \mathbf{h}=\mathbf{F d i a g}^{-1}(\mathbf{1}) \mathbf{F}^{H} \mathbf{y}=\mathbf{y} \rightarrow \mathbf{h}$ is the impulse response
- In the graph domain
- If we set $\mathbf{x}=\mathbf{e}_{i}$, then $\mathbf{h}=\boldsymbol{\Psi}^{-1} \operatorname{diag}^{-1}\left(\tilde{\mathbf{e}}_{i}\right) \mathbf{V}^{-1} \mathbf{y}$, where
$\Rightarrow \tilde{\mathbf{e}}_{i}:=\mathbf{V}^{-1} \mathbf{e}_{i} \equiv$ how strongly node $i$ expresses each of the freqs.
$\Rightarrow$ Problem if $\tilde{\mathbf{e}}_{i}$ has zero entries
- Alternatively we can get $\tilde{\mathbf{x}}=\mathbf{1}$ by setting $\mathbf{x}=\mathbf{V} 1$ and then

$$
\Rightarrow \mathbf{h}=\boldsymbol{\Psi}^{-1} \operatorname{diag}^{-1}(\tilde{\mathbf{x}}) \mathbf{V}^{-1} \mathbf{y}=\boldsymbol{\Psi}^{-1} \mathbf{V}^{-1} \mathbf{y}
$$

## Implementing graph filters: frequency or space

- Frequency or space?

$$
\mathbf{y}=\mathbf{V} \operatorname{diag}(\tilde{\mathbf{h}}) \mathbf{V}^{-1} \mathbf{x} \quad \text { vs. } \quad \mathbf{y}=\sum_{l=0}^{L} h_{l} \mathbf{S}^{\prime} \mathbf{x}
$$

- In space: leverage the fact that $\mathbf{S x}$ can be computed locally
$\Rightarrow$ Signal $\mathbf{x}$ is percolated $L$ times to find $\left\{\mathbf{x}^{(I)}\right\}_{l=0}^{L}$
$\Rightarrow$ Every node finds its own $y_{i}$ by computing $\sum_{l=0}^{L} h_{l}\left[\mathbf{x}^{(l)}\right]_{i}$
- Frequency implementation useful for processing if, e.g.,
$\Rightarrow$ Filter bandlimited and eigenvectors easy to find
$\Rightarrow$ Low complexity [Anis16, Tremblay16]
- Space definition useful for modeling
$\Rightarrow$ Diffusion, percolation, opinion formation, ... (more on this soon)
- More on filter design
$\Rightarrow$ Chebyshev polyn. [Shuman12]; AR-MA [Isufi-Leus15]; Node-var. [Segarra15]; Time-var. [Isufi-Leus16]; Median filters [Segarra16]


## Linear network processes via graph filters

- Consider a linear dynamics of the form

$$
\mathbf{x}_{t}-\mathbf{x}_{t-1}=\alpha \mathbf{J} \mathbf{x}_{t-1} \Rightarrow \mathbf{x}_{t}=(\mathbf{I}-\alpha \mathbf{J}) \mathbf{x}_{t-1}
$$

- If $\mathbf{x}$ is network process $\Rightarrow\left[\mathbf{x}_{t}\right]_{i}$ depends only on $\left[\mathbf{x}_{t-1}\right]_{j}, j \in \mathcal{N}(i)$


$$
[\mathrm{S}]_{i j}=[\mathrm{J}]_{i j} \Rightarrow \mathbf{x}_{t}=(\mathbf{I}-\alpha \mathbf{S}) \mathbf{x}_{t-1} \Rightarrow \mathbf{x}_{t}=(\mathbf{I}-\alpha \mathbf{S})^{t} \mathbf{x}_{0}
$$

$\Rightarrow \mathbf{x}_{t}=\mathbf{H} \mathbf{x}_{0}$, with $\mathbf{H}$ a polynomial of $\mathbf{S} \Rightarrow$ linear graph filter

- If the system has memory $\Rightarrow$ output weighted sum of previous exchanges (opinion dynamics) $\Rightarrow$ still a polynomial of $\mathbf{S}$

$$
\mathbf{y}=\sum_{t=0}^{T} \beta^{t} \mathbf{x}_{t} \Rightarrow \mathbf{y}=\sum_{t=0}^{T}(\beta \mathbf{I}-\beta \alpha \mathbf{S})^{t} \mathbf{x}_{0}
$$

- Everything holds true if $\alpha_{t}$ or $\beta_{t}$ are time varying


## Diffusion dynamics and AR (IIR) filters

- Before finite-time dynamics (FIR filters)
- Consider now a diffusion dynamics $\mathbf{x}_{t}=\alpha \mathbf{S} \mathbf{x}_{t-1}+\mathbf{w}$

$$
\begin{gathered}
\mathbf{x}_{t}=\alpha^{t} \mathbf{S}^{t} \mathbf{x}_{0}+\sum_{t^{\prime}=0}^{t} \alpha^{t} \mathbf{S}^{t^{\prime}} \mathbf{w} \\
\Rightarrow \text { When } t \rightarrow \infty: \quad \mathbf{x}_{\infty}=(\mathbf{I}-\alpha \mathbf{S})^{-1} \mathbf{w} \Rightarrow \text { AR graph filter }
\end{gathered}
$$



- Higher orders [Isufi-Leus16]
$\Rightarrow M$ successive diffusion dynamics $\Rightarrow A R$ of order $M$
$\Rightarrow$ Process is the sum of $M$ parallel diffusions $\Rightarrow$ ARMA order $M$

$$
\mathbf{x}_{\infty}=\prod_{m=1}^{M}\left(\mathbf{I}-\alpha_{m} \mathbf{S}\right)^{-1} \mathbf{w} \quad \mathbf{x}_{\infty}=\sum_{m=1}^{M}\left(\mathbf{I}-\alpha_{m} \mathbf{S}\right)^{-1} \mathbf{w}
$$

## General linear network processes

- Combinations of all the previous are possible

$$
\mathbf{x}_{t}=\mathbf{H}_{t}^{a}(\mathbf{S}) \mathbf{x}_{t-1}+\mathbf{H}_{t}^{b}(\mathbf{S}) \mathbf{w} \Rightarrow \mathbf{x}_{t}=\mathbf{H}_{t}^{A}(\mathrm{~S}) \mathbf{x}_{0}+\mathbf{H}_{t}^{B}(\mathrm{~S}) \mathbf{w}
$$

$\Rightarrow \mathbf{y}=\mathbf{x}_{t}$, sequential/parallel application, linear combination

$\Rightarrow$ Expands range of processes that can be modeled via GSP
$\Rightarrow$ Coefficients can change according to some control inputs

- A number of linear processes can be modeled using graph filters
$\Rightarrow$ Theoretical GSP results can be applied to distributed networking
$\Rightarrow$ Deconvolution, filtering, system id, ...
$\Rightarrow$ Beyond linearity possible too (more at the end of the talk)
- Links with control theory (of networks and complex systems)
$\Rightarrow$ Controllability, observability


## Application: Explaining human learning rates

- Why do some people learn faster than others?
$\Rightarrow$ Can we answer this by looking at their brain activity?
- Brain activity during learning of a motor skill in 112 cortical regions $\Rightarrow$ fMRI while learning a piano pattern for 20 individuals
- Pattern is repeated, reducing the time needed for execution
$\Rightarrow$ Learning rate $=$ rate of decrease in execution time
- Define a functional brain graph
$\Rightarrow$ Based on correlated activity
- fMRI outputs a series of graph signals
$\Rightarrow \mathbf{x}(t) \in \mathbb{R}^{112}$ describing brain states

- Does brain state variability correlate with learning?


## Measuring brain state variability

- We propose three different measures capturing different time scales $\Rightarrow$ Changes in micro, meso, and macro scales
- Micro: instantaneous changes higher than a threshold $\alpha$

$$
m_{1}(\mathbf{x})=\sum_{t=1}^{T} \mathbf{1}\left\{\frac{\|\mathbf{x}(t)-\mathbf{x}(t-1)\|_{2}}{\|\mathbf{x}(t)\|_{2}}>\alpha\right\}
$$

- Meso: Cluster brain states and count the changes in clusters

$$
m_{2}(\mathbf{x})=\sum_{t=1}^{T} \mathbf{1}\{\mathbf{c}(t) \neq \mathbf{c}(t-1)\}
$$

$\Rightarrow$ where $\mathbf{c}(t)$ is the cluster to which $\mathbf{x}(t)$ belongs.

- Macro: Sample entropy. Measure of complexity of time series

$$
\begin{aligned}
& \quad m_{3}(\mathbf{x})=-\log \left(\frac{\sum_{t} \sum_{s \neq t} \mathbf{1}\left\{\left\|\overline{\mathbf{x}}_{3}(t)-\overline{\mathbf{x}}_{3}(s)\right\|_{\infty}>\alpha\right\}}{\sum_{t} \sum_{s \neq t} \mathbf{1}\left\{\left\|\overline{\mathbf{x}}_{2}(t)-\overline{\mathbf{x}}_{2}(s)\right\|_{\infty}>\alpha\right\}}\right) \\
& \Rightarrow \text { Where } \overline{\mathbf{x}}_{r}(t)=[\mathbf{x}(t), \mathbf{x}(t+1), \ldots, \mathbf{x}(t+r-1)]
\end{aligned}
$$

## Diffusion as low-pass filtering

- We diffuse each time signal $\mathbf{x}(t)$ across the brain graph

$$
\mathbf{x}_{\mathrm{diff}}(t)=(\mathbf{I}+\beta \mathbf{L})^{-1} \mathbf{x}(t)
$$

$\Rightarrow$ where Laplacian $\mathbf{L}=\mathbf{V} \wedge \mathbf{V}^{-1}$ and $\beta$ represents the diffusion rate

- Analyzing diffusion in the frequency domain

$$
\begin{aligned}
& \quad \tilde{\mathbf{x}}_{\text {diff }}(t)=(\mathbf{I}+\beta \boldsymbol{\Lambda})^{-1} \mathbf{V}^{-1} \mathbf{x}(t)=\operatorname{diag}(\tilde{\mathbf{h}}) \tilde{\mathbf{x}}(t) \\
& \Rightarrow \text { where } \tilde{h}_{i}=1 /\left(1+\beta \lambda_{i}\right)
\end{aligned}
$$

- Diffusion acts as low-pass filtering
- High freq. components are attenuated
- $\beta$ controls the level of attenuation



## Computing correlation for three signals

- Variability measures consider the order of brain signal activity
- As a control, we include in our analysis a null signal time series $x_{\text {null }}$

$$
\mathbf{x}_{\mathrm{null}}(t)=\mathbf{x}_{\mathrm{diff}}\left(\pi_{t}\right)
$$

$\Rightarrow$ where $\pi_{t}$ is a random permutation of the time indices

- Correlation between variability ( $m_{1}, m_{2}$, and $m_{3}$ ) and learning?
- We consider three time series of brain activity
$\Rightarrow$ The original fMRI data x
$\Rightarrow$ The filtered data $\mathrm{x}_{\text {diff }}$
$\Rightarrow$ The null signal $\mathrm{x}_{\text {null }}$


## Low-pass filtering reveals correlation

- Correlation coeff. between learning rate and brain state variability

|  |  |  |  |
| :---: | :---: | :---: | :---: |
|  | Original | Filtered | Null |
| $m_{1}$ | 0.211 | $\mathbf{0 . 5 6 8}$ | 0.182 |
| $m_{2}$ | 0.226 | $\mathbf{0 . 6 1 1}$ | 0.174 |
| $m_{3}$ | 0.114 | $\mathbf{0 . 3 8 2}$ | 0.113 |

- Correlation is clear when the signal is filtered
$\Rightarrow$ Result for original signal similar to null signal
- Scatter plots for original, filtered, and null signals ( $m_{2}$ variability)





## Part II: Applications

## Motivation and preliminaries

Part I: Fundamentals
Graph signals and the shift operator
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Sampling graph signals Blind identification of graph filters
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Concluding remarks

## Application domains

- Design graph filters to approximate desired network operators
- Sampling bandlimited graph signals
- Blind graph filter identification
$\Rightarrow$ Infer diffusion coefficients from observed output
- Network topology inference
$\Rightarrow$ Infer shift from collection of network diffused signals

- Many more (not covered, glad to discuss or redirect):
$\Rightarrow$ Statistical GSP, stationarity and spectral estimation
$\Rightarrow$ Filter banks
$\Rightarrow$ Windowing, convolution, duality...
$\Rightarrow$ Nonlinear GSP


## Distributed network operators

- Design graph filters to implement a given linear transformation
$\Rightarrow$ Implementation is distributed by construction
$\Rightarrow$ Conditions for perfect and approximate implementation
$\Rightarrow$ [Shuman11], [Sandryhaila14], [Safavi15], [Chen15]
- Given a linear transformation B, find the filter coefficients h s. t.

$$
\mathbf{B}=\sum_{l=0}^{L-1} h_{l} \mathbf{S}^{\prime}
$$

$$
B=\left[\begin{array}{lll}
{\left[\begin{array}{ll}
1 \\
& \\
& \\
& \\
&
\end{array}\right]}
\end{array}\right.
$$

- Graph-shift operator S is given
$\Rightarrow$ Well-suited for cases where S is a network process
$\Rightarrow$ E.g., diffusion in a social network
$\Rightarrow$ Agents exchange information and weigh info observed
$\Rightarrow$ Choosing $\mathbf{h} \Rightarrow$ fixing the weights


## Conditions for perfect implementation

## Perfect implementation of linear graph operators [Segarra15]

The linear transformation B can be implemented using a graph filter $\mathbf{H}$ if the following conditions hold true:
i) Matrices B and S are simultaneously diagonalizable.
ii) If $\lambda_{k_{1}}=\lambda_{k_{2}}$, then $\gamma_{k_{1}}=\gamma_{k_{2}}$; and $L \geq \#\left\{\lambda_{k}\right\}_{k=1}^{N}$ distinct.

- i) $\Rightarrow$ frequency basis of $B$ and $S$ the same $\Rightarrow$ necessary
- ii) $\Rightarrow$ two equal freqs. in $S$ must be equal in $B \Rightarrow$ necessary
- Restrictive conditions but not impossible In time: to satisfy $\Rightarrow$ Consensus $\mathrm{B}_{\text {con }}=\mathbf{1 1}^{T}$
i) $\Leftrightarrow B$ favors i) and ii) because it is rank-one circulant
- If satisfied: $\mathbf{h}^{*}=\boldsymbol{\Psi}^{-1} \gamma$, where $\gamma=\left[\gamma_{1}, \ldots, \gamma_{N}\right]^{T}$ are eigvals. of $\mathbf{B}$


## Approximate design

- When perfect reconstruction is infeasible $\Rightarrow$ minimize error metric
$\Rightarrow$ Design $\mathbf{H x}$ to resemble Bx (or $\mathbf{H}$ to resemble $\mathbf{B}$ )
$\Rightarrow$ Minimizing $\left\|(\mathbf{H}-\mathbf{B}) \mathbf{R}_{\mathbf{x}}(\mathbf{H}-\mathbf{B})^{T}\right\|_{z}$ (with $\mathbf{R}_{x}=\mathbf{I}$ if unknown)
- MSE coefficients: $\mathbf{h}^{*}=\Theta_{\mathbf{R}_{x}}^{\dagger} \mathbf{b}_{\mathbf{R}_{x}}=\left(\Theta_{\mathbf{R}_{x}}^{T} \Theta_{\mathbf{R}_{x}}\right)^{-1} \Theta_{\mathbf{R}_{x}}^{T} \mathbf{b}_{\mathbf{R}_{x}}$
$\Rightarrow$ with $\boldsymbol{\Theta}_{\mathbf{R}_{\mathrm{x}}}:=\left[\operatorname{vec}\left(\mathbf{I}_{\mathrm{x}}^{1 / 2}\right), \ldots, \operatorname{vec}\left(\mathbf{S}^{L-1} \mathbf{R}_{\mathrm{x}}^{1 / 2}\right)\right], \mathbf{b}_{\mathbf{R}_{\mathrm{x}}}:=\operatorname{vec}\left(\mathbf{B R}_{\mathrm{x}}^{1 / 2}\right)$
- Worst-case error coefficients:

$$
\begin{aligned}
& \left\{\mathbf{h}^{*}, s^{*}\right\}=\underset{\{\mathbf{h}, s\}}{\operatorname{argmin}} s \\
& \text { s. to }\left[\begin{array}{cc}
s \mathbf{I} \\
\left(\mathbf{V} \operatorname{diag}(\Psi \mathbf{h}) \mathbf{V}^{-1}-\mathbf{B}\right)^{T} & \mathbf{V} \operatorname{diag}(\Psi \mathbf{h}) \mathbf{V}^{-1}-\mathbf{B} \\
s \mathbf{R}_{\mathbf{x}}^{-1}
\end{array}\right] \succeq 0 .
\end{aligned}
$$

- Additional assumptions can be incorporated


## Consensus and rank- 1 transformations

## Consensus

- Local implementation of the consensus operator $\mathrm{B}_{\mathrm{con}}=\mathbf{1 1}^{T} / \mathrm{N}$


Proposition [Segarra16]
If $\mathcal{G}$ is connected and the desired operator $\mathrm{B}_{\mathrm{rk} 1}$ is rank one, then there exists an $\mathbf{S}$ such that $\mathbf{B}_{\mathrm{rk} 1}$ can be written as a graph filter $\sum_{l=0}^{N-1} h_{l} \mathbf{S}^{\prime}$.

- Constructive proof, for consensus $\mathbf{S}=\mathbf{L}$
- Consensus is achieved in finite time [Sandryhaila-Kar-Moura14]
- Key: B low-rank (repeated eigenvalues) $\Rightarrow$ well-suited for approx.
- We compare the performance of: 1) Asymptotic fastest distributed linear averaging (FDLA), 2) Graph filter approx.


## Finite-time consensus

- Define the graph-shift operator $\mathbf{S}=\mathrm{W}$
$\Rightarrow$ Where $\lim _{k \rightarrow \infty} \mathbf{W}^{k}=\mathbf{B}_{\text {con }}$ with fastest convergence
- Plot average errors across the 100 graphs with 10 nodes
- Compare worst-case and mean error design (50 nodes)


- Smaller error than FDLA for intermediate $K$
$\Rightarrow$ When $K=N-1=9$, perfect recovery
- The price to pay is that $\left\{\lambda_{k}\right\}_{k=1}^{N}$ need to be known
- Consistent performance of mean error and worst case designs


## Node-variant graph filters: definition

- A generalization of graph filters [Segarra16]:

$$
\mathbf{H}_{\mathrm{nv}}:=\sum_{l=0}^{L-1} \operatorname{diag}\left(\mathbf{h}^{(/)}\right) \mathbf{S}^{\prime}
$$

$\Rightarrow$ When $\mathbf{h}^{(I)}=h_{l} \mathbf{1} \Rightarrow$ regular (node-invariant) filter


$$
\mathbf{x}^{(0)}=\mathbf{x}=\left(\begin{array}{r}
-1 \\
2 \\
0 \\
0 \\
0 \\
0
\end{array}\right) \quad \mathbf{x}^{(1)}=\mathbf{S} \mathbf{x}^{(0)}=\left(\begin{array}{r}
2 \\
-1 \\
2 \\
0 \\
1 \\
0
\end{array}\right) \quad \mathbf{x}^{(2)}=\mathbf{S} \mathbf{x}^{(1)}=\left(\begin{array}{r}
0 \\
3 \\
-1 \\
3 \\
1 \\
0
\end{array}\right)
$$

$$
\mathbf{y}=h^{(0)} \mathbf{x}^{(0)}+h^{(1)} \mathbf{x}^{(1)}+h^{(2)} \mathbf{x}^{(2)}
$$

$$
y_{i}=h_{i}^{(0)} x_{i}^{(0)}+h_{i}^{(1)} x_{i}^{(1)}+h_{i}^{(2)} x_{i}^{(2)}
$$

- In general, when $\mathbf{H}_{\mathrm{nv}}$ is applied to a signal $\mathbf{x}$
$\Rightarrow$ Each node applies different weights to the shifted signals $\mathbf{S}^{\prime} \mathbf{x}$
$\Rightarrow$ More flexible and still distributed, not shift-invariant


## Node-variant graph filters: frequency response

- Collect the coefficients of node $i$ in $\mathbf{h}_{i}$, such that $\left[\mathbf{h}_{i}\right]_{I}=\left[\mathbf{h}^{(/)}\right]_{i}$
- Focus on the filter output at node $i, \mathbf{e}_{i}^{T} \mathbf{H}_{\mathrm{nv}} \mathbf{x}$

$$
\boldsymbol{\eta}_{i}^{T}=\mathbf{e}_{i}^{T} \mathbf{H}_{\mathrm{nv}}=\sum_{l=0}^{L-1}\left[\mathbf{h}_{i}\right] / \mathbf{e}_{i}^{T} \mathbf{V} \Lambda^{\prime} \mathbf{V}^{-1}
$$



- Defining $\mathbf{u}_{i}:=\mathbf{V}^{\top} \mathbf{e}_{i}$

$$
\boldsymbol{\eta}_{i}^{T}=\mathbf{u}_{i}^{T}\left(\sum_{l=0}^{L-1}\left[\mathbf{h}_{i}\right] / \boldsymbol{\Lambda}^{\prime}\right) \mathbf{V}^{-1}=\mathbf{u}_{i}^{T} \operatorname{diag}\left(\boldsymbol{\Psi} \mathbf{h}_{i}\right) \mathbf{V}^{-1}
$$

- The output of the filter at node $i, \boldsymbol{\eta}_{i}^{T} \mathbf{x}$ is the inner product of $\Rightarrow \mathbf{V}^{-1} \mathbf{x} \Rightarrow$ the frequency representation of the input, and $\Rightarrow \mathbf{u}_{i} \Rightarrow$ how strongly the frequencies are expressed by node $i$ $\Rightarrow$ Modulated by $\boldsymbol{\psi} \mathbf{h}_{i} \Rightarrow$ Frequency response associated to $i$


## Perfect reconstruction with node-variant filters

- Node-variant filters can implement a large class of transformations $\Rightarrow \operatorname{Pick} \mathbf{h}^{(I)}$ for $I=0, \cdots, L-1$ so that $\mathbf{B}=\sum_{l=0}^{L-1} \operatorname{diag}\left(\mathbf{h}^{(I)}\right) \mathbf{S}^{\prime}$
$\Rightarrow \mathrm{TH}$ : Always possible if $\mathbf{V}$ non-zero and $\left\{\lambda_{k}\right\}$ distinct
- Application in distributed processing: analog network coding $\Rightarrow \mathbf{B}$ is a binary matrix (input-output pairs)
- Example: $G$ undirected, with $N=10, \mathbf{S}=\mathbf{A}$, sources 3 and 6 $\Rightarrow$ Node 3 tx to 1, 4, 6, 7, and 10; node 6 to the remaining ones
$\Rightarrow$ Node invariant unable to implement B

$t=0$

$t=1$

$t=2$

$t=3$



## Sampling bandlimited graph signals

## Motivation and preliminaries

Part I: Fundamentals
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## Motivation and preliminaries

- Sampling and interpolation are cornerstone problems in classical SP
$\Rightarrow$ How recover a signal using only a few observations?
$\Rightarrow$ Need to limit the degrees of freedom: subspace, smoothness
- Graph signals: sampling thoroughly investigated
$\Rightarrow$ Most works assume only a few values are observed
$\Rightarrow$ [Anis14, Chen15, Tsitsvero15, Puy15, Wang15]

- Alternative approach [Marques16, Segarra16]
$\Rightarrow$ GSP is well-suited for distributed networking
$\Rightarrow$ Incorporate local graph structure into the observation model
$\Rightarrow$ Recover signal using distributed local graph operators




## Sampling bandlimited graph signals: Overview

- Sampling is likely to be most important inverse problem
$\Rightarrow$ How to find $\mathbf{x} \in \mathbb{R}^{N}$ using $P<N$ observations?
- Our focus on bandlimited signals, but other models possible
$\Rightarrow \tilde{\mathbf{x}}=\mathbf{V}^{-1} \mathbf{x}$ sparse
$\Rightarrow \mathbf{x}=\sum_{k \in \mathcal{K}} \tilde{x}_{k} \mathbf{v}_{k}$, with $|\mathcal{K}|=K<N$
$\Rightarrow \mathbf{S}$ involved in generation of $\mathbf{x}$
$\Rightarrow$ Agnostic to the particular form of $\mathbf{S}$

- Two sampling schemes were introduced in the literature
$\Rightarrow$ Selection [Anis14, Chen15, Tsitsvero15, Puy15, Wang15]
$\Rightarrow$ Aggregation [Segarra15], [Marques15]
$\Rightarrow$ Hybrid scheme combining both $\Rightarrow$ Space-shift sampling
- More involved, theoretical benefits, practical benefits in distr. setups


## Revisiting sampling in time

- There are two ways of interpreting sampling of time signals
- We can either freeze the signal and sample values at different times

- We can fix a point (present) and sample the evolution of the signal

- Both strategies coincide for time signals but not for general graphs $\Rightarrow$ Give rise to selection and aggregation sampling


## Selection sampling: Definition

- Intuitive generalization to graph signals

$$
\Rightarrow \mathbf{C} \in\{0,1\}^{P \times N} \text { (matrix } P \text { rows of } \mathbf{I}_{N} \text { ) }
$$

$\Rightarrow$ Sampled signal is $\overline{\mathbf{x}}=\mathbf{C x}$


- Goal: recover x based on $\overline{\mathrm{x}}$
$\Rightarrow$ Assume that the support of $\mathcal{K}$ is known (w.l.o.g. $\mathcal{K}=\{k\}_{k=1}^{K}$ )
$\Rightarrow$ Since $\tilde{x}_{k}=0$ for $k>K$, define $\tilde{\mathbf{x}}_{K}:=\left[\tilde{x}_{1}, \ldots, \tilde{x}_{K}\right]^{T}=\mathbf{E}_{K}^{T} \tilde{\mathbf{x}}$

- Approach: use $\overline{\mathbf{x}}$ to find $\tilde{\mathbf{x}}_{K}$, and then recover $\mathbf{x}$ as

$$
\mathbf{x}=\mathbf{V}\left(\mathbf{E}_{K} \tilde{\mathbf{x}}_{K}\right)=\left(\mathbf{V E} \mathbf{E}_{K}\right) \tilde{\mathbf{x}}_{K}=\mathbf{V}_{K} \tilde{\mathbf{x}}_{K}
$$

## Selection sampling: Recovery

- Number of samples $P \geq K$

$$
\overline{\mathbf{x}}=\mathbf{C} \mathbf{x}=\mathbf{C} \mathbf{V}_{K} \tilde{\mathbf{x}}_{K}
$$

$\Rightarrow\left(\mathbf{C V}_{K}\right)$ submatrix of $\mathbf{V}$


Recovery of selection sampling
If $\operatorname{rank}\left(\mathbf{C V}_{K}\right) \geq K, \mathbf{x}$ can be recovered from the $P$ values in $\overline{\mathbf{x}}$ as

$$
\mathbf{x}=\mathbf{V}_{K} \tilde{\mathbf{x}}_{K}=\mathbf{V}_{K}\left(\mathbf{C} \mathbf{V}_{K}\right)^{\dagger} \overline{\mathbf{x}}
$$

- With $P=K$, hard to check invertibility (by inspection)
$\Rightarrow$ Columns of $\mathbf{V}_{K}\left(\mathbf{C} \mathbf{V}_{K}\right)^{-1}$ are the interpolators
- In time ( $\mathbf{S}=\mathbf{A}_{d c}$ ), if the samples in $\mathbf{C}$ are equally spaced
$\Rightarrow\left(\mathbf{C V} \mathbf{V}_{K}\right)$ is Vandermonde (DFT) and $\mathbf{V}_{K}\left(\mathbf{C} \mathbf{V}_{K}\right)^{-1}$ are sincs


## Aggregation sampling: Definition

- Idea: incorporating $\mathbf{S}$ to the sampling procedure
$\Rightarrow$ Reduces to classical sampling for time signals
- Consider shifted (aggregated) signals $\mathbf{y}^{(1)}=\mathbf{S}^{\prime} \mathbf{x}$
$\Rightarrow \mathbf{y}^{(I)}=\mathbf{S} \mathbf{y}^{(I-1)} \Rightarrow$ found sequentially with only local exchanges
- Form $\mathbf{y}_{i}=\left[y_{i}^{(0)}, y_{i}^{(1)}, \ldots, y_{i}^{(N-1)}\right]^{T}$ (obtained locally by node $i$ )

- The sampled signal is

$$
\overline{\mathbf{y}}_{i}=\mathbf{C y} \mathbf{y}_{i}
$$

- Goal: recover $\mathbf{x}$ based on $\overline{\mathbf{y}}_{i}$


## Aggregation sampling: Recovery

- Goal: recover $\mathbf{x}$ based on $\overline{\mathbf{y}}_{i} \Rightarrow$ Same approach than before $\Rightarrow$ Use $\overline{\mathbf{y}}_{i}$ to find $\tilde{\mathbf{x}}_{K}$, and then recover $\mathbf{x}$ as $\mathbf{x}=\mathbf{V}_{K} \tilde{\mathbf{x}}_{K}$
- Define $\overline{\mathbf{u}}_{i}:=\mathbf{V}_{K}^{T} \mathbf{e}_{i}$ and recall $\Psi_{k l}=\lambda_{k}^{I-1}$


## Recovery of aggregation sampling

Signal $\mathbf{x}$ can be recovered from the first $K$ samples in $\overline{\mathbf{y}}_{i}$ as

$$
\mathbf{x}=\mathbf{V}_{K} \tilde{\mathbf{x}}_{K}=\mathbf{V}_{K} \operatorname{diag}^{-1}\left(\overline{\mathbf{u}}_{i}\right)\left(\mathbf{C} \boldsymbol{\Psi}^{T} \mathbf{E}_{K}\right)^{-1} \overline{\mathbf{y}}_{i}
$$

provided that $\left[\overline{\mathbf{u}}_{i}\right]_{k} \neq 0$ and all $\left\{\lambda_{k}\right\}_{k=1}^{K}$ are distinct.

- If $\mathbf{C}=\mathbf{E}_{K}^{T}$, node $i$ can recover $\mathbf{x}$ with info from $K-1$ hops!
$\Rightarrow$ Node $i$ has to be able to capture frequencies in $\mathcal{K}$
$\Rightarrow$ The frequencies have to distinguishable
- Bandlimited signals: Signals that can be well estimated locally


## Aggregation and selection sampling: Example

- In time $\left(\mathbf{S}=\mathbf{A}_{d c}\right)$, selection and aggregation are equivalent
$\Rightarrow$ Differences for a more general graph?
- Erdős-Rényi
$p=0.2, \mathbf{S}=\mathbf{A}$,
$K=3$, non-smooth

- First 3 observations at node 4: $\mathbf{y}_{4}=[0.55,1.27,2.94]^{T}$
$\Rightarrow\left[\mathbf{y}_{4}\right]_{1}=x_{4}=-0.55,\left[\mathbf{y}_{4}\right]_{2}=x_{2}+x_{3}+x_{5}+x_{6}+x_{7}=1.27$
$\Rightarrow$ For this example, any node guarantees recovery
$\Rightarrow$ Selection sampling fails if, e.g., $\{1,3,4\}$


## Sampling: Discussion and extensions

- Discussion on aggregation sampling
$\Rightarrow$ Observation matrix: diagonal times Vandermonde
$\Rightarrow$ Very appropriate in distributed scenarios
$\Rightarrow$ Different nodes will lead to different performance (soon)
$\Rightarrow$ Types of signals that are actually bandlimited (role of $\mathbf{S}$ )
- Three extensions:
$\Rightarrow$ Sampling in the presence of noise
$\Rightarrow$ Unknown frequency support
$\Rightarrow$ Space-shift sampling (hybrid)


## Presence of noise

- Linear observation model: $\overline{\mathbf{z}}_{i}=\mathbf{C} \boldsymbol{\Psi}_{i} \tilde{\mathbf{x}}_{K}+\mathbf{C} \mathbf{w}_{i}$ and $\mathbf{x}=\mathbf{V}_{K} \tilde{\mathbf{x}}_{K}$
- BLUE interpolation ( $\boldsymbol{\Psi}_{i}$ either selection or aggregation)

$$
\begin{aligned}
& \hat{\tilde{\mathbf{x}}}_{K}^{(i)}=\left[\boldsymbol{\Psi}_{i}^{H} \mathbf{C}^{H}\left(\overline{\mathbf{R}}_{w}^{(i)}\right)^{-1} \mathbf{C} \boldsymbol{\Psi}_{i}\right]^{-1} \boldsymbol{\Psi}_{i}^{H} \mathbf{C}^{H}\left(\overline{\mathbf{R}}_{w}^{(i)}\right)^{-1} \overline{\mathbf{z}}_{i} \\
& \Rightarrow \text { If } P=K \text {, then } \hat{\mathbf{x}}^{(i)}=\mathbf{V}_{K}\left(\mathbf{C} \boldsymbol{\Psi}_{i}\right)^{-1} \overline{\mathbf{z}}_{i}
\end{aligned}
$$

- Error covariances $\left(\mathbf{R}_{e}^{(i)}, \tilde{\mathbf{R}}_{e}^{(i)}\right)$ in closed form $\Rightarrow$ Noise covariances?
$\Rightarrow$ Colored, different models: white noise in $\mathbf{z}_{i}$, in $\mathbf{x}$, or in $\tilde{\mathbf{x}}_{K}$
- Metric to optimize?

$$
\Rightarrow \operatorname{trace}\left(\mathbf{R}_{e}^{(i)}\right), \lambda_{\max }\left(\mathbf{R}_{e}^{(i)}\right), \log \operatorname{det}\left(\tilde{\mathbf{R}}_{e}^{(i)}\right),\left[\operatorname{trace}\left(\tilde{\mathbf{R}}_{e}^{(i)^{-1}}\right)\right]^{-1}
$$

- Select $i$ and $\mathbf{C}$ to min. error $\Rightarrow$ Depends on metric and noise [Marques16]


## Unknown frequency support

- Falls into the class of sparse reconstruction: observation matrix?
$\Rightarrow$ Selec. $\Rightarrow$ submatrix of unitary $\mathbf{V}_{\mathcal{K}}$
$\Rightarrow$ Aggr. $\Rightarrow$ Vander. $\times$ diag

$$
\left[\mathbf{u}_{i}\right]_{k} \neq 0 \text { and } \lambda_{k} \neq \lambda_{k^{\prime}} \Rightarrow \text { full-spark }
$$



- Joint recovery and support identification (noiseless)

$$
\begin{aligned}
& \tilde{\mathbf{x}}^{*}:= \arg \min _{\tilde{\mathbf{x}}} \quad\|\tilde{\mathbf{x}}\|_{0} \\
& \text { s.t. } \quad \mathbf{C y}_{i}=\mathbf{C} \Psi_{i} \tilde{\mathbf{x}},
\end{aligned}
$$

- If full spark $\Rightarrow P=2 K$ samples suffice
$\Rightarrow$ Different relaxations are possible
$\Rightarrow$ Conditioning will depend on $\boldsymbol{\Psi}_{i}$ (e.g., how different $\left\{\lambda_{k}\right\}$ are)
- Noisy case: sampling nodes critical


## Recovery with unknown support: Example

- Erdős-Rényi

$$
p=0.15,0.20,0.25
$$

$K=3$, non-smooth


- Three different shifts: $\mathbf{A},(\mathbf{I}-\mathbf{A})$ and $\frac{1}{2} \mathbf{A}^{2}$





## Space-shift sampling

- Space-shift sampling (hybrid) $\Rightarrow$ Multiple nodes and multiple shifts

Selection: 4 nodes, 1 sample


Space-shift: 2 nodes, 2 samples


Aggregat.: 1 node, 4 samples


- Section and aggregation sampling as particular cases
- With $\overline{\mathbf{U}}:=\left[\operatorname{diag}\left(\overline{\mathbf{u}}_{1}\right), \ldots, \operatorname{diag}\left(\overline{\mathbf{u}}_{N}\right)\right]^{T}$, the sampled signal is

$$
\underline{\underline{\mathbf{z}}}=\mathbf{C}\left(\mathbf{I} \otimes\left(\boldsymbol{\Psi}^{T} \mathbf{E}_{K}\right)\right) \overline{\mathbf{U}}_{\tilde{\mathbf{x}}}^{K}+\mathbf{C} \underline{\mathbf{w}}
$$

- As before, BLUE and error covariance in close-form
- Optimizing sample selection more challenging
- More structured schemes easier: e.g., message passing
$\Rightarrow$ Node $i$ knows $y_{i}^{(I)} \Rightarrow$ node $i$ knows $y_{j}^{\left(I^{\prime}\right)}$ for all $j \in \mathcal{N}_{i}$ and $I^{\prime}<I$


## Sampling the US economy

- 62 economic sectors in USA +2 artificial sectors
$\Rightarrow$ Graph: average flows in 2007-2010, $\mathbf{S}=\mathbf{A}$
$\Rightarrow$ Signal x: production in 2011
$\Rightarrow \mathbf{x}$ is approximately bandlimited with $K=4$





## Sampling the US economy: Results

- Setup 1: we add different types of noise
$\Rightarrow$ Error depends on sampling node: better if more connected


- Setup 2: we try different shift-space strategies

|  | Sampling strategy |  |  | Min. error | Median error |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $[\mathbf{x}]_{i}$ | $[\mathbf{S x}]_{i}$ | $\left[\mathbf{S}^{\mathbf{2}} \mathbf{x}\right]_{i}$ | $\left[\mathbf{S}^{\mathbf{3}} \mathbf{x}\right]_{i}$ | .0035 | .019 |
| $[\mathbf{x}]_{i}$ | $[\mathbf{x}]_{j}$ | $[\mathbf{x}]_{k}$ | $[\mathbf{x}]_{l}$ | .0039 | 4.2 |
| $[\mathbf{S x}]_{i}$ | $[\mathbf{S x}]_{j}$ | $[\mathbf{S x}]_{k}$ | $[\mathbf{S x}]_{l}$ | .0035 | .030 |
| $\left[\mathbf{S}^{\mathbf{2}} \mathbf{x}\right]_{i}$ | $\left[\mathbf{S}^{\mathbf{2}} \mathbf{x}\right]_{j}$ | $\left[\mathbf{S}^{\mathbf{2}} \mathbf{x}\right]_{k}$ | $\left[\mathbf{S}^{2} \mathbf{x}\right]_{l}$ | .0035 | .0055 |
| $\left[\mathbf{S}^{\mathbf{3}} \mathbf{x}\right]_{i}$ | $\left[\mathbf{S}^{\mathbf{3}} \mathbf{x}\right]_{j}$ | $\left[\mathbf{S}^{\mathbf{3}} \mathbf{x}\right]_{k}$ | $\left[\mathbf{S}^{\mathbf{3}} \mathbf{x}\right]_{l}$ | .0035 | .0040 |
| $[\mathbf{x}]_{i}$ | $[\mathbf{S x}]_{i}$ | $[\mathbf{x}]_{j}$ | $[\mathbf{S x}]_{j}$ | .0035 | .039 |

## More on sampling graph signals

- Beyond bandlimitedness
$\Rightarrow$ Smooth signals [Chen15]
$\Rightarrow$ Parsimonious in kernelized domain [Romero-Giannakis16]
- Strategies to select the sampling nodes
$\Rightarrow$ Random (sketching) [Varma15]
$\Rightarrow$ Optimal reconstruction [Marques16, Chepuri-Leus16]
$\Rightarrow$ Designed based on posterior task [Gama16]
- And more...
$\Rightarrow$ Low-complexity implementations [Tremblay16, Anis16]
$\Rightarrow$ Local implementations [Wang14, Segarra15]
$\Rightarrow$ Unknown spectral decomposition [Anis16]


## Blind identification of graph filters

## Motivation and preliminaries

Part I: Fundamentals
Graph signals and the shift operator
Graph Fourier Transform (GFT)
Graph filters and network processes
Part II: Applications
Filter design for network operators
Sampling graph signals Blind identification of graph filters
Network topology inference
Concluding remarks

## Diffusion processes as graph filter outputs

- Q: Upon observing a graph signal $\mathbf{y}$, how was this signal generated?
- Postulate the following generative model
$\Rightarrow$ An originally sparse signal $x=\mathbf{x}^{(0)}$
$\Rightarrow$ Diffused via linear graph dynamics $\mathbf{S} \Rightarrow \mathbf{x}^{(I)}=\mathbf{S} \mathbf{x}^{(I-1)}$
$\Rightarrow$ Observed $\mathbf{y}$ is a linear combination of the diffused signals $\mathbf{x}^{(1)}$

$$
\mathbf{y}=\sum_{l=0}^{L} h_{l} \mathbf{x}^{(l)}=\sum_{l=0}^{L} h_{l} \mathbf{S}^{\prime} \mathrm{x}=\mathrm{H} \mathbf{x}
$$

- Model: Observed network process as output of a graph filter $\Rightarrow$ View few elements in $\operatorname{supp}(\mathrm{x})=:\left\{i: x_{i} \neq 0\right\}$ as seeds


## Motivation and problem statement

- Ex: Global opinion/belief profile formed by spreading a rumor
$\Rightarrow$ What was the rumor? Who started it?
$\Rightarrow$ How do people weigh in peers' opinions to form their own?

- Problem: Blind identification of graph filters with sparse inputs
- Q: Given S, can we find $x$ and the combination weights $h$ from $y=H x$ ?
$\Rightarrow$ Extends classical blind deconvolution to graphs


## Blind graph filter identification

- Leverage frequency response of graph filters ( $\mathbf{U}:=\mathbf{V}^{-1}$ )

$$
\mathbf{y}=\mathbf{H x} \Rightarrow \mathbf{y}=\mathbf{V} \operatorname{diag}(\boldsymbol{\Psi} \mathbf{h}) \mathbf{U} \mathbf{x}
$$

$\Rightarrow \mathbf{y}$ is a bilinear function of the unknowns $\mathbf{h}$ and $\mathbf{x}$

- Problem is ill-posed $\Rightarrow(L+1)+N$ unknowns and $N$ observations
$\Rightarrow$ As.: x is $S$-sparse i.e., $\|\mathbf{x}\|_{0}:=|\operatorname{supp}(\mathbf{x})| \leq S$
- Blind graph filter identification $\Rightarrow$ Non-convex feasibility problem

$$
\text { find }\{\mathbf{h}, \mathbf{x}\}, \quad \text { s. to } \mathbf{y}=\mathbf{V} \operatorname{diag}(\boldsymbol{\Psi} \mathbf{h}) \mathbf{U} \mathbf{x},\|\mathbf{x}\|_{0} \leq S
$$

## "Lifting" the bilinear inverse problem

- Key observation: Use the Khatri-Rao product $\odot$ to write $\mathbf{y}$ as

$$
\mathbf{y}=\mathbf{V}\left(\Psi^{T} \odot \mathbf{U}^{T}\right)^{T} \operatorname{vec}\left(\mathbf{x h}^{T}\right)
$$

- Reveals $\mathbf{y}$ is a linear combination of the entries of $\mathbf{Z}:=\mathbf{x h}^{\top}$

- Z is of rank- 1 and row-sparse $\Rightarrow$ Linear matrix inverse problem

$$
\min _{\mathbf{Z}} \operatorname{rank}(\mathbf{Z}), \quad \text { s. to } \mathbf{y}=\mathbf{V}\left(\boldsymbol{\Psi}^{T} \odot \mathbf{U}^{T}\right)^{T} \operatorname{vec}(\mathbf{Z}), \quad\|\mathbf{Z}\|_{2,0} \leq S
$$

$\Rightarrow$ Pseudo-norm \|Z $\|_{2,0}$ counts the nonzero rows of $\mathbf{Z}$
$\Rightarrow$ Matrix "lifting" for blind deconvolution [Ahmed etal'14]

- Rank minimization s. to row-cardinality constraint is NP-hard. Relax!


## Algorithmic approach via convex relaxation

- Key property: $\ell_{1}$-norm minimization promotes sparsity [Tibshirani'94]
- Nuclear norm $\|\mathbf{Z}\|_{*}:=\sum_{i} \sigma_{i}(\mathbf{Z})$ a convex proxy of rank [Fazel'01]
- $\ell_{2,1}$ norm $\|Z\|_{2,1}:=\sum_{i}\left\|\mathbf{z}_{i}^{T}\right\|_{2}$ surrogate of $\|\mathbf{Z}\|_{2,0}$ [Yuan-Lin'06]
- Convex relaxation

$$
\min _{\mathbf{Z}}\|\mathbf{Z}\|_{*}+\alpha\|\mathbf{Z}\|_{2,1}, \quad \text { s. to } \mathbf{y}=\mathbf{V}\left(\boldsymbol{\Psi}^{T} \odot \mathbf{U}^{T}\right)^{T} \operatorname{vec}(\mathbf{Z})
$$

$\Rightarrow$ Scalable algorithm using method of multipliers

- Refine estimates $\{\mathbf{h}, \mathbf{x}\}$ via iteratively-reweighted optimization
$\Rightarrow$ Weights $\alpha_{i}(k)=\left(\left\|\mathbf{z}_{i}(k)^{T}\right\|_{2}+\delta\right)^{-1}$ per row $i$, per iteration $k$
- Noisy and partial observations $\Rightarrow$ Adjust constraints
- Noise in $\mathbf{y}:\left\|\mathbf{y}-\mathbf{V}\left(\boldsymbol{\Psi}^{T} \odot \mathbf{U}^{T}\right)^{T} \operatorname{vec}(\mathbf{Z})\right\| \leq \varepsilon$
- Sampling via selection matrix $\mathbf{C}: \mathbf{y}_{C}=\mathbf{C V}\left(\boldsymbol{\Psi}^{T} \odot \mathbf{U}^{T}\right)^{T} \operatorname{vec}(Z)$


## Exact recovery guarantees

- Exact recovery $\Rightarrow$ Success of the convex relaxation
- Random model on the graph structure [Ling-Stromher'15]
- Probabilistic guarantees depend on graph spectrum

$$
P_{\mathrm{rec}} \geq 1-N^{-O\left(\rho_{\mathbf{u}}^{-1}(S)\right)}, \quad \rho_{\mathbf{u}}(S):=\max _{I \in\{1, \ldots, N\}} \max _{\Omega \in \Omega_{S}^{N}}\left\|\mathbf{u}_{l, \Omega}\right\|_{2}^{2}
$$



- Blind deconvolution (in time) most favorable graph setting


## Numerical tests: Recovery rates

- Recovery rates over an $(L, S)$ grid and 20 trials
- Successful recovery when $\left\|\mathbf{x}^{*}\left(\mathbf{h}^{*}\right)^{T}-\mathbf{x h}^{T}\right\|_{\mathrm{F}}<10^{-3}$
- ER (left), ER reweighted $\ell_{2,1}$ (center), brain reweighted $\ell_{2,1}$ (right)



- Exact recovery over non-trivial $(L, S)$ region
$\Rightarrow$ Reweighted optimization markedly improves performance
$\Rightarrow$ Encouraging results even for real-world graphs


## Numerical tests: Brain graph

- Human brain graph with $N=66$ regions, $L=3$ and $S=3$

- Proposed method also outperforms alternating-minimization solver
$\Rightarrow$ Unknown $\operatorname{supp}(\mathbf{x}) \approx$ Need twice as many observations
$\Rightarrow$ Stable to Gaussian noise in $\mathbf{y}\left(\sigma^{2}=0.01\right)$


## Multiple output signals

- Suppose we have access to $P$ output signals $\left\{\mathbf{y}_{p}\right\}_{p=1}^{P}$

- Goal: Identify common filter $\mathbf{H}$ fed by multiple unobserved inputs $\mathrm{x}_{p}$


## Formulation

- As.: $\left\{\mathrm{x}_{p}\right\}_{p=1}^{P}$ are $S$-sparse with common support
- Concatenate outputs $\overline{\mathbf{y}}:=\left[\mathbf{y}_{1}^{T}, \ldots, \mathbf{y}_{P}^{T}\right]^{T}$ and inputs $\overline{\mathrm{x}}:=\left[\mathrm{x}_{1}^{T}, \ldots, \mathrm{x}_{P}^{T}\right]^{T}$
- Unknown rank-one matrices $Z_{p}:=x_{p} \mathbf{h}^{T}$. Stack them
$\Rightarrow$ Vertically in rank one $\overline{\mathbf{Z}}_{v}:=\left[\mathbf{Z}_{1}^{T}, \ldots, \mathbf{Z}_{P}^{T}\right]^{T}=\overline{\mathrm{x}} \mathrm{h}^{T} \in \mathbb{R}^{N P \times L}$
$\Rightarrow$ Horizontally in row sparse $\overline{\mathbf{Z}}_{h}:=\left[\mathbf{Z}_{1}, \ldots, \mathrm{Z}_{P}\right] \in \mathbb{R}^{N \times P L}$
- Convex formulation

$$
\begin{aligned}
& \min _{\left\{\mathbf{Z}_{p}\right\}_{p=1}^{P}}\left\|\overline{\mathbf{Z}}_{v}\right\|_{*}+\tau\left\|\overline{\mathbf{Z}}_{h}\right\|_{2,1}, \quad \text { s. to } \overline{\mathbf{y}}=\left(\mathbf{I}_{P} \otimes\left(\mathbf{V}\left(\Psi^{T} \odot \mathbf{U}^{T}\right)^{T}\right)\right) \operatorname{vec}\left(\overline{\mathbf{Z}}_{h}\right) \\
& \Rightarrow \operatorname{Relax}(\mathbf{A s .}):\left\|\overline{\mathbf{Z}}_{h}\right\|_{2,1} \leftrightarrow\left\|\overline{\mathbf{Z}}_{v}\right\|_{2,1}=\sum_{p=1}^{P}\left\|\mathbf{Z}_{p}\right\|_{2,1}
\end{aligned}
$$

## Numerical tests: Multiple signals, recovery rates

- Recovery rates over an $(L, S)$ grid and 20 trials
- Successful recovery when $\left\|\hat{\overline{\mathbf{x}}} \hat{\mathbf{h}}^{T}-\overline{\mathbf{x}} \mathbf{h}^{T}\right\|_{\mathrm{F}}<10^{-3}$
- ER (left), ER reweighted $\ell_{2,1}$ (center), brain reweighted $\ell_{2,1}$ (right)







- Leveraging multiple output signals aids the blind identification task


## Blind ID: Takeaways

- Extended blind deconvolution of space/time signals to graphs
$\Rightarrow$ Key: model diffusion process as output of graph filter
- Rank and sparsity minimization subject to model constraints
$\Rightarrow$ "Lifting" and convex relaxation yield efficient algorithms
- Exact recovery conditions $\Rightarrow$ Success of the convex relaxation $\Rightarrow$ Probabilistic guarantees that depend on the graph spectrum
- Consideration of multiple sparse inputs aids recovery
- Envisioned application domains
(a) Opinion formation in social networks
(b) Identify sources of epileptic seizure
(c) Trace "patient zero" for an epidemic outbreak
- Unknown shift S $\Rightarrow$ Network topology inference


## Network topology inference

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## Motivation and context

- Network topology inference from nodal observations
$\Rightarrow$ Approaches use Pearson correlations to construct graphs
$\Rightarrow$ Partial correlations and conditional dependence
- Paramount importance in neuroscience
$\Rightarrow$ Functional net inferred from activity

- Most GSP works assume that $\mathbf{S}$ (hence the graph) is known
$\Rightarrow$ Analyze how the characteristics of $\mathbf{S}$ affect signals and filters
- We take the reverse path
$\Rightarrow$ How to use GSP to infer the graph topology?
$\Rightarrow$ [Dong15, Mei15, Pavez16, Pasdeloup16]


## Generating structure of a diffusion process

- Signal $\mathbf{x}$ is the response of a linear diffusion process to a white input

$$
\mathbf{x}=\alpha_{0} \prod_{l=1}^{\infty}\left(\mathbf{I}-\alpha_{l} \mathbf{S}\right) \mathbf{w}=\sum_{l=0}^{\infty} \beta_{l} \mathbf{S}^{\prime} \mathbf{w}
$$

$\Rightarrow$ Common generative model. Heat diffusion if $\alpha_{k}$ constant

- We say the graph shift $\mathbf{S}$ explains the structure of signal x
- It follows from Cayley Hamilton that we can write diffusion as

$$
\mathbf{x}=\left(\sum_{l=0}^{N-1} h_{l} \mathbf{S}^{\prime}\right) \mathbf{w}:=\mathbf{H} \mathbf{w}
$$

$\Rightarrow \mathbf{H}$ diagonalized by the eigenvectors of the shift operator

## Our approach for topology inference

- We propose a two-step approach for graph topology identification

- Beyond diffusion $\Rightarrow$ alternative sources for spectral templates $\mathbf{V}$


## STEP 1: Obtaining the eigenvectors

- The covariance matrix of the signal $\mathbf{x}$ is

$$
\mathbf{C}_{x}=\mathbb{E}\left[\left(\mathbf{H} \mathbf{w}(\mathbf{H} \mathbf{w})^{H}\right)\right]=\mathbf{H} \mathbb{E}\left[\left(\mathbf{w} \mathbf{w}^{H}\right)\right] \mathbf{H}^{H}=\mathbf{H} \mathbf{H}^{H}
$$

- Since $\mathbf{H}$ is diagonalized by $\mathbf{V}$, so is the covariance $\mathbf{C}_{x}$

$$
\mathbf{C}_{x}=\mathbf{V}\left|\sum_{l=0}^{L-1} h_{l} \boldsymbol{\Lambda}^{\prime}\right|^{2} \mathbf{V}^{H}=\mathbf{V} \operatorname{diag}\left(|\tilde{\mathbf{h}}|^{2}\right) \mathbf{V}^{H}
$$

- Any shift with eigenvectors $\mathbf{V}$ can explain $\mathbf{x}$
$\Rightarrow \mathrm{G}$ and its specific eigenvalues have been obscured by diffusion
Observations
(a) There are many shifts that can explain a signal $\mathbf{x}$
(b) Identifying the shift $\mathbf{S}$ is just a matter of identifying the eigenvalues
(c) In correlation methods the eigenvalues are kept unchanged
(d) In precision methods the eigenvalues are inverted


## Other sources of spectral templates

1) Implementation of linear network operators

- Goal: distributed implementation of linear operator B via graph filter $\Rightarrow \mathbf{B}$ and $\mathbf{S}$ sharing $\mathbf{V}$ is beneficial for implementation
- Given a pre-specified B
$\Rightarrow$ Use its eigenvectors as spectral templates to generate a shift S
$\Rightarrow$ The goal here not to identify a shift, but to design one
Ex.: consensus $\Rightarrow$ Laplacian of the smallest connected graph

2) Relationship between nodes of a signal

- Particular transforms T are known to work well on specific data
$\Rightarrow$ Such transform assumes an implicit relation among data $\Rightarrow \mathbf{S}$
$\Rightarrow$ Identification of that relation can provide insights $\mathbf{V}^{H}=\mathbf{T}$

DCTs: i-iii


## STEP 2: Obtaining the eigenvalues

- We can use extra knowledge/assumptions to choose one graph
$\Rightarrow$ Of all graphs, select one that is optimal in some sense

$$
\begin{equation*}
\mathbf{S}^{*}:=\underset{\mathbf{s}, \boldsymbol{\lambda}}{\operatorname{argmin}} f(\mathbf{S}, \boldsymbol{\lambda}) \quad \text { s. to } \quad \mathbf{S}=\sum_{k=1}^{N} \lambda_{k} \mathbf{v}_{k} \mathbf{v}_{k}^{H}, \mathbf{S} \in \mathcal{S} \tag{1}
\end{equation*}
$$

- Set $\mathcal{S}$ contains all admissible scaled adjacency matrices

$$
\mathcal{S}:=\left\{\mathbf{S} \mid S_{i j} \geq 0, \quad \mathbf{S} \in \mathcal{M}^{N}, S_{i i}=0, \quad \sum_{j} S_{1 j}=1\right\}
$$

$\Rightarrow$ Can accommodate Laplacian matrices as well

- Problem is convex if we select a convex objective $f(\mathbf{S}, \boldsymbol{\lambda})$
$\Rightarrow$ Minimum energy $\left(f(\mathbf{S})=\|\mathbf{S}\|_{F}\right)$, Fast mixing $\left(f(\boldsymbol{\lambda})=-\lambda_{2}\right)$


## Size of the feasibility set

- The feasibility set in (1) is generally small
$\Rightarrow$ Define $\mathbf{W}:=\mathbf{V} \odot \mathbf{V}$ where $\odot$ is the Khatri-Rao product
$\Rightarrow$ Denote by $\mathcal{D}$ the index set such that $\operatorname{vec}(\mathbf{S})_{\mathcal{D}}=\operatorname{diag}(\mathbf{S})$

Assume that (1) is feasible, then it holds that $\operatorname{rank}\left(\mathbf{W}_{\mathcal{D}}\right) \leq N-1$. If $\operatorname{rank}\left(\mathrm{W}_{\mathcal{D}}\right)=N-1$, then the feasible set of $(1)$ is a singleton.

- Convex feasibility set $\Rightarrow$ Search for the optimal solution may be easy
- Simulations will show that $\operatorname{rank}\left(\mathbf{W}_{\mathcal{D}}\right)=N-1$ arises in practice


## Sparse recovery

- Whenever the feasibility set of (1) is non-trivial
$\Rightarrow f(\mathbf{S}, \boldsymbol{\lambda})$ determines the features of the recovered graph
Ex: Identify the sparsest shift $\mathbf{S}_{0}^{*}$ that explains observed signal structure $\Rightarrow$ Set the cost $f(\mathbf{S}, \boldsymbol{\lambda})=\|\mathbf{S}\|_{0}$
- Problem is not convex, but can relax to $\ell_{1}$ norm minimization

$$
\mathbf{S}_{1}^{*}:=\underset{\mathbf{s}, \boldsymbol{\lambda}}{\operatorname{argmin}}\|\mathbf{S}\|_{1} \quad \text { s. to } \quad \mathbf{S}=\sum_{k=1}^{N} \lambda_{k} \mathbf{v}_{k} \mathbf{v}_{k}^{H}, \mathbf{S} \in \mathcal{S}
$$

- Does the solution $\mathbf{S}_{1}^{*}$ coincide with the $\ell_{0}$ solution $\mathbf{S}_{0}^{*}$ ?


## Recovery guarantee

- Denoting by $\mathbf{m}_{i}^{T}$ the $i$-th row of $\mathbf{M}:=\left(\mathbf{I}-\mathbf{W} \mathbf{W}^{\dagger}\right)_{\mathcal{D}^{c}}$
$\Rightarrow$ Construct R:= $\left[\mathbf{m}_{2}-\mathbf{m}_{1}, \ldots, \mathbf{m}_{N-1}-\mathbf{m}_{1}, \mathbf{m}_{N}, \ldots, \mathbf{m}_{\left|\mathcal{D}^{c}\right|}\right]^{T}$
$\Rightarrow$ Denote by $\mathcal{K}$ the indices of the support of $\mathrm{s}_{0}^{*}=\operatorname{vec}\left(\mathbf{S}_{0}^{*}\right)$
$\mathbf{S}_{1}^{*}$ and $\mathbf{S}_{0}^{*}$ coincide if the two following conditions are satisfied:

1) $\operatorname{rank}\left(\mathrm{R}_{\mathcal{K}}\right)=|\mathcal{K}|$; and
2) There exists a constant $\delta>0$ such that

$$
\psi_{\mathbf{R}}:=\left\|\mathbf{I}_{\mathcal{K}^{c}}\left(\delta^{-2} \mathbf{R} \mathbf{R}^{T}+\mathbf{I}_{\mathcal{K}^{c}}^{T} \mathbf{I}_{\mathcal{K}^{c}}\right)^{-1} \mathbf{I}_{\mathcal{K}}^{T}\right\|_{\infty}<1 .
$$

- Cond. 1) ensures uniqueness of solution $\mathbf{S}_{1}^{*}$
- Cond. 2) guarantees existence of a dual certificate for $\ell_{0}$ optimality


## Noisy and incomplete spectral templates

- We might have access to $\hat{\mathbf{V}}$, a noisy version of the spectral templates
$\Rightarrow$ With $d(\cdot, \cdot)$ denoting a (convex) distance between matrices

$$
\min _{\{\mathbf{S}, \lambda, \hat{\mathbf{S}}\}}\|\mathbf{S}\|_{1} \quad \text { s. to } \hat{\mathbf{S}}=\sum_{k=1}^{N} \lambda_{k} \hat{\mathbf{v}}_{k} \hat{\mathbf{v}}_{k}^{T}, \quad \mathbf{S} \in \mathcal{S}, \quad d(\mathbf{S}, \hat{\mathbf{S}}) \leq \epsilon
$$

- Recovery result similar to the noiseless case can be derived $\Rightarrow$ Conditions under which we are guaranteed $d\left(\mathbf{S}^{*}, \mathbf{S}_{0}^{*}\right) \leq C \epsilon$
- Partial access to $\mathrm{V} \Rightarrow$ Only $K$ known eigenvectors $\left[v_{1}, \ldots, v_{K}\right.$ ]

$$
\min _{\left\{\mathbf{S}, \mathbf{S}_{\bar{K}}, \boldsymbol{\lambda}\right\}}\|\mathbf{S}\|_{1} \text { s. to } \mathbf{S}=\mathbf{S}_{\bar{K}}+\sum_{k=1}^{K} \lambda_{k} \mathbf{v}_{k} \mathbf{v}_{k}^{T}, \mathbf{S} \in \mathcal{S}, \quad \mathbf{S}_{\bar{K}} \mathbf{v}_{k}=\mathbf{0}
$$

- Incomplete and noisy scenarios can be combined


## Topology inference in random graphs

- Erdős-Rényi graphs of varying size $N \in\{10,20, \ldots, 50\}$
$\Rightarrow$ Edge probabilities $p \in\{0.1,0.2, \ldots, 0.9\}$
- Recovery rates for adjacency (left) and normalized Laplacian (mid)



- Recovery is easier for intermediate values of $p$
- Rate of recovery related to the $\operatorname{rank}\left(\mathbf{W}_{\mathcal{D}}\right)$ (histogram $N=10, p=0.2$ )
$\Rightarrow$ When rank is $N-1$, recovery is guaranteed
$\Rightarrow$ As rank decreases, there is a detrimental effect on recovery


## Sparse recovery guarantee

- Generate 1000 ER random graphs ( $N=20, p=0.1$ ) such that $\Rightarrow$ Feasible set is not a singleton
$\Rightarrow$ Cond. 1) in sparse recovery theorem is satisfied
- Noiseless case: $\ell_{1}$ norm guarantees recovery as long as $\psi_{\mathbf{R}}<1$

- Condition is sufficient but not necessary
$\Rightarrow$ Tightest possible bound on this matrix norm


## Inferring brain graphs from noisy templates

- Identification of structural brain graphs $N=66$
- Test recovery for noisy spectral templates $\hat{\mathbf{V}}$
$\Rightarrow$ Obtained from sample covariances of diffused signals

- Recovery error decreases with increasing number of observed signals
$\Rightarrow$ More reliable estimate of the covariance $\Rightarrow$ Less noisy $\hat{\mathbf{V}}$
- Brain of patient 1 is consistently the hardest to identify
$\Rightarrow$ Robustness for identification in noisy scenarios
- Traditional methods like graphical lasso fail to recover S


## Inferring social graphs from incomplete templates

- Identification of multiple social networks $N=32$
$\Rightarrow$ Defined on the same node set of students from Ljubljana
- Test recovery for incomplete spectral templates $\hat{\mathbf{V}}=\left[\mathbf{v}_{1}, \ldots, \mathbf{v}_{K}\right]$
$\Rightarrow$ Obtained from a low-pass diffusion process
$\Rightarrow$ Repeated eigenvalues in $\mathbf{C}_{\mathbf{x}}$ introduce rotation ambiguity in $\mathbf{V}$

- Recovery error decreases with increasing nr. of spectral templates
$\Rightarrow$ Performance improvement is sharp and precipitous


## Performance comparisons

- Comparison with graphical lasso and sparse correlation methods
- Evaluated on 100 realizations of ER graphs with $N=20$ and $p=0.2$

- Graphical lasso implicitly assumes a filter $\mathbf{H}_{1}=(\rho \mathbf{I}+\mathbf{S})^{-1 / 2}$
$\Rightarrow$ For this filter spectral templates work, but not as well (MLE)
- For general diffusion filters $\mathbf{H}_{2}$ spectral templates still work fine


## Inferring direct relations

- Our method can be used to sparsify a given network
- Keep direct and important edges or relations
$\Rightarrow$ Discard indirect relations that can be explained by direct ones
- Use eigenvectors $\hat{V}$ of given network as noisy templates
- Infer contact between amino-acid residues in BPT1 BOVIN
$\Rightarrow$ Use mutual information of amino-acid covariation as input


Ground truth


Mutual info.


Network deconv.


Our approach

- Network deconvolution assumes a specific filter model [Feizi13]
$\Rightarrow$ We achieve better performance by being agnostic to this


## Topology ID: Takeaways

- Network topology inference cornerstone problem in Network Science
- Most GSP works analyze how $\mathbf{S}$ affect signals and filters
- Here, reverse path: How to use GSP to infer the graph topology?
- Our GSP approach to network topology inference
$\Rightarrow$ Two step approach: i) Obtain $\mathbf{V}$; ii) Estimate $\mathbf{S}$ given $\mathbf{V}$
- How to obtain the spectral templates $\mathbf{V}$
$\Rightarrow$ Based on covariance of diffused signals
$\Rightarrow$ Other sources too: net operators, data transforms
- Infer S via convex optimization
$\Rightarrow$ Objectives promotes desirable properties
$\Rightarrow$ Constraints encode structure a priori info and structure
$\Rightarrow$ Formulations for perfect and imperfect templates
$\Rightarrow$ Sparse recovery results for both adjacency and Laplacian


## Wrapping up

## Motivation and preliminaries

Part I: Fundamentals
Graph signals and the shift operator
Graph Fourier Transform (GFT)
Graph filters and network processes
Part II: Applications
Filter design for network operators
Sampling graph signals Blind identification of graph filters
Network topology inference
Concluding remarks

## Concluding remarks

- Network science and big data pose new challenges
$\Rightarrow$ GSP can contribute to solve some of those challenges
$\Rightarrow$ Well suited for network (diffusion) processes
- Central elements in GSP: graph-shift operator and Fourier transform
- Graph filters: operate graph signals
$\Rightarrow$ Polynomials of the shift operator that can be implemented locally
- Network diffusion/percolations processes via graph filters
$\Rightarrow$ Successive/parallel combination of local linear dynamics
$\Rightarrow$ Possibly time-varying diffusion coefficients
$\Rightarrow$ Accurate to model certain setups
$\Rightarrow$ GSP yields insights on how those processes behave


## Concluding remarks

- GSP results can be applied to solve practical problems
$\Rightarrow$ Filter design (design of distributed operators)
$\Rightarrow$ Sampling, interpolation (network control)
$\Rightarrow$ Blind deconvolution (source ID), shift design (network topology ID)

Interpolate a brain signal from local observations


Smooth an observed network profile

Compress a signal in
an irregular domain


Predict the evolution of a network process

Localize the source of a rumor


Infer the topology where the signals reside

## Looking ahead

- First step to challenging problems: social nets, brain signals
- Motivates further research:
$\Rightarrow$ Statistical modeling
$\Rightarrow$ Space-time variation
$\Rightarrow$ Changing topologies
$\Rightarrow$ Nonlinear approaches
$\Rightarrow$ Local, reduced-complexity algorithms
- Thanks!
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$\Rightarrow$ Slides on stationarity available at:
http://tsc.urjc.es/~amarques/papers/ssamglar_sam16_slides.pdf


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